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FREQUENCY METHODS IN COMPUTER AIDED DESIGN OF CONTROL SYSTEMS

STEPHEN CHRISTIE SPIELMAN

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FREQUENCY METHODS IN COMPUTER AIDED DESIGN OF CONTROL SYSTEMS

BY

STEPHEN CHRISTIE SPIELMAN

B.S., University of Illinois, 1974

THESIS

Submitted in partial fulfillment of the requirements
for the degree of Master of Science in Electrical Engineering
in the Graduate College of the
University of Illinois at Urbana-Champaign, 1977

Thesis Adviser: Professor William R. Perkins

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FREQUENCY METHODS IN COMPUTER AIDED
DESIGN OF CONTROL SYSTEMS

by

Stephen Christie Spielman

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CHAPTER 1

AN ALTERNATIVE APPROACH

State space methods have served an important role in the development of modern control theory. They were originally proposed to help solve the increasingly complex aerospace control problems which frequency response methods of the time could not deal with. Historically this theory has lead to a much greater understanding of the general nature of a system and to the solution of a great variety of control problems. These means have further been applied to many other areas in the sciences, thus firmly establishing the importance of modern system studies.

While the elegance of these methods cannot be denied, for the linear and time invariant system they pose many practical difficulties. These include incorporation of simplicity, low sensitivity, and robustness of the resulting control, as well as integrity and fault tolerance, in a design. Working with an input-output description of the plant is believed to solve many of these problems [1].

In fact, the classical frequency domain design of single input single output systems has never been replaced by state space methods in practice.

Rosenbrock et al., at the Control Systems Centre, University of Manchester, England, has extended these frequency domain techniques to multiple input multiple output systems. This design method, which he has named the inverse Nyquist array method, encompasses a large number of control problems. One of its fortes is the ability of the designer to shape the outcome of the resulting designs to suit his needs and still achieve an adequate system performance.

How this is possible will become evident in what follows.

Chapter 2 relies heavily on Rosenbrock's own exposition of the method [2].

Chapter 3 presents an outline of programs to be used for interactive frequency domain design with computer graphics facilities. An example is given in Chapter 4.

CHAPTER 2

THE INVERSE NYQUIST ARRAY METHOD

2.1. Introducing Specifications and the Plant

The design process begins with an input-output description of the plant. Let the plant be linear and time invariant. It has an $(\ell \times 1)$ vector \bar{u} as input, and an $(m \times 1)$ vector \bar{y} as output. In the design, one is primarily interested in the response at each output of the plant as we apply sinusoidal forcing functions of various frequencies to each input. Therefore the plant is represented by an $m \times \ell$ transfer function matrix $\overline{G(s)}$, which is rational, with the exception of the inclusion of pure exponential terms of the form e^{sT} . These terms represent pure time delay and may be included when necessary, as many plants exhibit this sort of behavior.

It is possible that a state space representation of the plant, arising from a physical model or available empirical data, is at hand. This, of course, is a suitable description of the plant after the required transfer functions are obtained via analysis. As one's actual interest is in the complex frequency mapping between input and output, it may be more appropriate to measure this data on the actual plant, if the physical situation allows this. For the type plants being considered the data may always be fitted into such a transfer function matrix expression. While this is not implicitly a necessary task to perform, it shall be assumed that the transfer function matrix is of the above form. In the case of designing from actual measured data, $\overline{G(s)}$ merely represents this data and is a convenient means to interpolate between the data points for the physical model at hand.

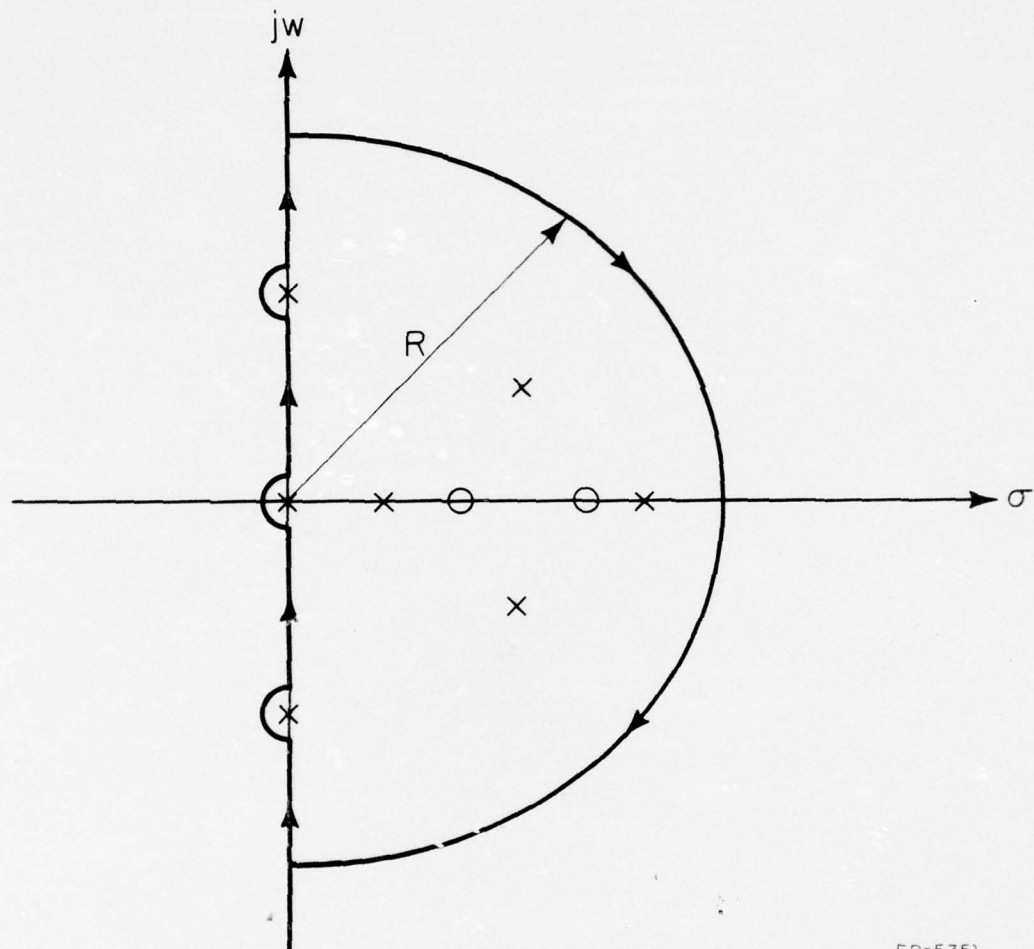
As usual in frequency domain techniques, one deals with the map of the open loop transfer function in the complex plane as a suitably defined contour is traversed. The encirclements of a critical point then indicate stability. Here the contour, which shall always be denoted by D , is in every case traversed clockwise. It consists of that segment of the imaginary axis from $-jR$ to $+jR$, as well as the semicircle of radius R about the origin in the right half plane. The contour is indented to the left of any pole or zero of the transfer function at the origin or on the $j\omega$ axis, as one in general is interested in asymptotically stable systems and, therefore, must include these points within the closed path. The radius R is taken to be large enough to insure that all right half plane poles and zeros are enclosed by D . See Figure 2.1.1.

Given a $\overline{G(s)}$, consider a design procedure based on the closed loop configuration shown in Figure 2.1.2. An $(l \times k)$ input compensation matrix $\overline{K(s)}$ relates the error \overline{E} with input \overline{U} . Similarly an $(k \times m)$ output compensator matrix $\overline{L(s)}$ relates output \overline{Y} to \overline{Z} , while a $(k \times k)$ feedback matrix $\overline{F(s)}$ is inserted in the return path.

For notational simplicity, the $(k \times k)$ open loop transfer function matrix of the system relating the error $\overline{E(s)}$ to the output $\overline{Z(s)}$ is defined as \overline{Q}

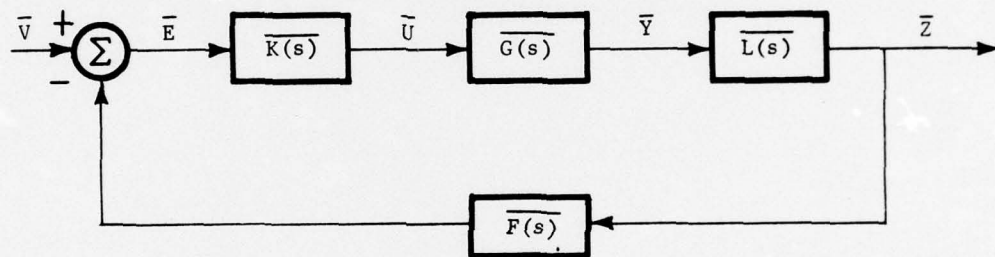
$$\overline{Q(s)} = \overline{L(s)} \overline{G(s)} \overline{K(s)}. \quad (2.1.1)$$

As one is interested in the closed loop responses, the $(k \times k)$ closed loop transfer function matrix relating inputs $\overline{V(s)}$ to outputs $\overline{Z(s)}$ is defined as $\overline{H(s)}$. By assuming



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Figure 2.2.1. The Nyquist contour in the complex plane.



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Figure 2.1.2. System block diagram.

$$|\bar{I}_k + \overline{Q(s)} \overline{F(s)}| \neq 0 \quad (2.1.2)$$

it immediately follows that

$$\bar{H}(s) = (\bar{I}_k + \overline{Q(s)} \overline{F(s)})^{-1} \overline{Q(s)}. \quad (2.1.3)$$

In assuming

$$|\bar{I}_k + \overline{F(s)} \overline{Q(s)}| \neq 0 \quad (2.1.4)$$

the closed loop transfer function may be expressed as

$$\overline{H(s)} = \overline{Q(s)} (\bar{I}_k + \overline{F(s)} \overline{Q(s)})^{-1}. \quad (2.1.5)$$

As both (2.1.3) and (2.1.5) show that \bar{H} is a function of \bar{Q} , \bar{F} , and the Laplace variable s , one might consider

$$\bar{H} = \bar{H}[\bar{Q}, \bar{F}, s] \quad (2.1.6)$$

as a compact way of expressing this closed loop transfer function matrix. While this represents a relatively simple expression in terms of computational considerations, it shows very little in terms of relating the open loop system to the closed loop system. As our interest is in designing for an acceptable closed loop system performance, a further investigation into a possible means to these ends becomes necessary.

2.2. Inverse Relationships and their Consequences

The results in the previous section are valid provided that the determinants (2.1.2) and (2.1.4) exist. It should now be assumed that

$$|\bar{Q}| \neq 0 \quad (2.2.1)$$

which implies

$$|\bar{K}| \neq 0, \quad |\bar{G}| \neq 0 \quad |\bar{L}| \neq 0. \quad (2.2.2)$$

If this were not the case, the system would not be functionally controllable and therefore, be of very little practical interest. Functional controllability, also known as functional reproducibility, loosely means that there exists some input which will generate a specific bounded output from zero initial conditions, for all time other than the initial time [3, p. 169].

Provided that (2.2.1) is satisfied, \bar{Q} will have an inverse which shall be denoted by a circumflex.

$$\bar{Q}^{-1} = \hat{Q} = (\bar{L} \bar{G} \bar{K})^{-1} = \hat{K} \hat{G} \hat{L} \quad (2.2.3)$$

A similar notation shall be used for all other matrix inverses. Note that the (i,j) element of \hat{Q} is referred to as \hat{q}_{ij} , whereas the inverse of that same element in Q is referred to as $(q_{ij})^{-1}$.

Using inverse matrices, the closed loop transfer function matrix implicitly defined in (2.1.6) has an inverse of

$$\hat{H} = \bar{F} + \hat{Q}. \quad (2.2.4)$$

This shows that if \hat{Q} exists, the closed loop inverse transfer function can be expressed in the astoundingly simple form of (2.2.4). Being additive by nature, it is the convenient means of relating the closed loop system to the open loop system which is desired. Clearly, more insight can be found in working with the inverse system and the easy transition in closing the loops, than from working with the noninversed system and dealing with the nonintuitive nature of the matrix multipliers imbedded in (2.1.6).

In a design one is interested in controlling the k outputs \bar{Z} with the respective k inputs \bar{V} . To do this, the open loop inverse transfer function \hat{Q} is constructed so as to minimize the effect of the off diagonal entries in \bar{G} . Once a sufficient degree of noninteraction is present, the original problem has been considerably reduced.

It is, of course, always possible to construct \hat{Q} so that the desired noninteraction is present simply by choosing \hat{K} or \hat{L} equal to \bar{G} . This will make \hat{Q} diagonal and equal to the identity matrix. While such a method has been suggested, it imposes severe problems in terms of plant variations, implementation in terms of stable subsystems, and the resulting shape of the frequency loci [4]. Perhaps the most objectionable reason is the necessary complexity of the resulting control scheme. As simplicity in a design often makes the difference between an engineering success and a failure, one could never justify such a scheme as this with its inherent faults.

Quite the contrary, one in fact wishes to find a matrix compensator which is as simple as possible in order to achieve a required amount of noninteractiveness. Ideally, this compensating matrix would be independent of frequency and as sparse as possible. This is the easiest control to implement on the actual plant, and would probably be the most reliable as well.

When sufficient decoupling of the open loop system is attained, appropriate stability regions can be determined so that feedback may be applied. By considering

$$\bar{F} = \text{diag}\{f_i\}, \quad i = 1, 2, \dots, k \quad (2.2.5)$$

where each f_i is scalar and independent of s , the closed loop frequency loci of the diagonal transfer functions in \hat{H} become those of the corresponding loci in \hat{Q} with the imaginary axis appropriately shifted. Single loop frequency compensation in the classical sense of phase-lag and phase-lead networks may be designed to modify these frequency loci for the desired input-output responses. In short, our multivariable control problem becomes k single loop designs, in which an abundance of frequency domain techniques exist [2, pp. 28-116], and more modern algorithmic methods have never entirely replaced.

Before showing how this is possible, it is necessary to formalize the intuitive idea of noninteractiveness. As the design means adopted are graphical in nature, this information must be included on the diagonal inverse frequency loci of each individual control loop.

2.3. Dominant Systems

Rosenbrock calls these loosely coupled or noninteractive systems dominant, as the principle diagonal element primarily determines that input-output relationship.

2.3.1. Diagonal Dominance

Definition 1 [2, p. 142]: A complex rational $(k \times k)$ matrix $\overline{Z(s)}$ is said to be diagonally row dominant on the contour D if $z_{ii}(s)$ has no pole on D for all

$i = 1, 2, \dots, k$, and

$$|z_{ii}(s)| - \sum_{\substack{j=1 \\ j \neq i}}^k |z_{ij}(s)| > 0 \quad \forall i = 1, 2, \dots, k \quad (2.3.1)$$

and for all s on D .

Definition 2 [2, p. 142]: A complex rational $(k \times k)$ matrix $\overline{Z(s)}$ is said to be diagonally column dominant on the contour D if $z_{ii}(s)$ has no pole on D for all $i = 1, 2, \dots, k$, and

$$|z_{ii}(s)| - \sum_{\substack{j=1 \\ j \neq i}}^k |z_{ji}(s)| > 0 \quad \forall i = 1, 2, \dots, k \quad (2.3.2)$$

and all s on D .

Definition 3 [2, p. 142]: $\overline{Z(s)}$ is diagonally dominant on D if it is either diagonally row dominant or diagonally column dominant.

Observe that row dominance implies dominance, and column dominance implies dominance. Then for dominance, the sum of the moduli of the off diagonal row elements or column elements must be greater than the modules of that diagonal element for all s on D . It is clear that rows and columns may not be mixed to check dominance at a particular s on D . Instead, one may have either row dominance, or column dominance as s takes on different values around D , and still satisfy Definition 3.

A simple graphical construction exists to check the dominance of $\hat{Q}(s)$. Let $\hat{q}_{ii}(s)$ map the familiar Nyquist contour D into Γ_i . For a particular s on D , draw a circle with its center at the corresponding point on Γ_i of radius

$$d_i(s) = \sum_{\substack{j=1 \\ j \neq i}}^k |\hat{q}_{ij}(s)| \quad (2.3.3)$$

which is based on rows, or

$$d'_i(s) = \sum_{\substack{j=1 \\ j \neq i}}^k |\hat{q}_{ji}(s)| \quad (2.3.4)$$

which is based on columns. Do this for a sufficient number of s on D so as to define an area about Γ_i . If the area defined by this set of circles of radius d_i excludes the origin, for all $i=1,2,\dots,k$, then \hat{Q} is row dominant on D . Similarly if the area defined by the set of circles of radius d'_i excludes the origin for all $i=1,2,\dots,k$, then Q is column dominant on D . In this same sense, if none of the circles of either radius d_i or of radius d'_i has the origin as an interior point or on its circumference, for all i as before, then \hat{Q} is dominant on D . These bands, as defined by (2.3.2) and (2.3.3) are called Gershgorin bands as their usefulness is dependent upon Gershgorin's Theorem.*

2.3.2. Ostrowski's Theorem

Systems which are diagonally dominant have determinants which show interesting properties and will be of use later in investigating stability. However, the following theorem, due to Ostrowski [5], which was rediscovered by Rosenbrock, allows for the location of the diagonal transfer function in a dominant system.

* Gershgorin's Theorem, while being of little practical value for this design method, has important consequences in the rigorous development of what follows. For its statement and proof, one is referred to the literature [3, p. 11].

Theorem 1: Ostrowski's Theorem [2,p. 149]: Let the rational complex ($k \times k$) matrix $\overline{Z}(s)$ be row (or respectively column) dominant for $s = s_0$ on any simple closed contour C , having on it no pole of any $z_{ii}(s)$. Then $\overline{Z}(s_0)$ has an inverse $\hat{Z}(s_0)$ which, for all $i = 1, 2, \dots, k$ satisfies:

$$|\hat{z}_{ii}(s_0)^{-1} - z_{ii}(s_0)| < \phi_i(s_0) d_i(s_0) < d_i(s_0) \quad (2.3.5)$$

where

$$\phi_i(s_0) = \max_{j \neq i} d_j(s_0) / |z_{jj}(s_0)| \quad (2.3.6)$$

and $d_i(s_0)$ is given by (2.3.3) for row dominance., or alternatively, which satisfies

$$|z_{ii}(s_0)^{-1} - z_{ii}(s_0)| < \phi'_i(s_0) d'_i(s_0) < d'_i(s_0) \quad (2.3.7)$$

where

$$\phi'_i(s_0) = \max_{j \neq i} d'_j(s_0) / |z_{jj}(s_0)| \quad (2.3.8)$$

and $d'_i(s_0)$ is given by (2.3.4), for column dominance.

The proof of Ostrowski's theorem may be found elsewhere [3,pp. 204-206]. This theorem of course has several important design implications.

Theorem 2 [2, p. 150]: Let \hat{Q} exist and satisfy the conditions of Ostrowski's Theorem. Furthermore, assume \bar{F} is diagonal and independent of s . Then for each s on D the diagonal element h_{ii} of (2.1.6) satisfies

$$|h_{ii}^{-1}(s) - (f_i + \hat{q}_{ii}(s))| < \phi_i(s) d_i(s) < d_i(s) \quad (2.3.9)$$

if \hat{H} is column dominant, or

$$|h_{ii}^{-1}(s) - (f_i + \hat{q}_{ii}(s))| < \phi'_i(s) d'_i(s) < d'_i(s) \quad (2.3.10)$$

if \hat{H} is row dominant.

The proof of this theorem follows as a direct consequence of Ostrowski's Theorem. Consider $\bar{Z} = \hat{H} = \bar{F} + \hat{Q}$, which makes $\hat{z}_{ii}^{-1} = h_{ii}^{-1}$ and $z_{ii} = \hat{h}_{ii} = f_i + \hat{q}_{ii}$. Then

$$\phi_i(s_o) = \max_{\substack{j \\ j \neq i}} \frac{d_j(s_o)}{|f_j + \hat{q}_{jj}(s_o)|} \quad (2.3.11)$$

if \hat{H} is column dominant, or

$$\phi'_i(s_o) = \max_{\substack{j \\ j \neq i}} \frac{d'_j(s_o)}{|f_j + \hat{q}_{jj}(s_o)|} \quad (2.3.12)$$

if \hat{H} is row dominant, and the desired result is obtained. The set of circles defined by (2.3.11) and (2.3.12) lie inside the corresponding Gershgorin bands. They are called Ostrowski bands in honor of the mathematician on whose theorem they are based. Theorem 2 has the following graphical consequences. Pick a particular input-output relationship characterized by a diagonal element in \hat{H} , say \hat{h}_{jj} . If the gains

$$\{f_1, f_2, \dots, f_{j-1}, f_{j+1}, \dots, f_k\} \quad (2.3.13)$$

are kept constant while allowing f_j to vary, then one has essentially a single input single output situation. It is exactly this cause and effect relationship for which we wish to design a single loop controller. Then Theorem 2 tells us that the inverse transfer function h_{ii}^{-1} lies somewhere within the set of Ostrowski bands based on the constant gains (2.3.13). Furthermore, one may vary f_j and thus design for the appropriate stability in terms of gain margins, phase margins, other Nyquist diagram type

specifications [2, pp. 33-35] and their corresponding time domain interpretations, from the inverse Nyquist diagram with the appropriate Ostrowski bands. When these bands are small enough, the resulting composite diagram is treated as if it were just a "fuzzy" inverse Nyquist diagram

2.4. Obtaining Dominance

In Figure 2.1.2 the forward path transfer function matrix Q was forced to be square to allow the possible existence of its necessary inverse. This was done by making matrices \bar{K} and \bar{L} of the appropriate dimensions. It now becomes necessary to make the further assumption that the plant transfer function matrix \bar{G} is square. This implies that \bar{K} and \bar{L} are also square or $k = l = m$.

As an aside, it should be mentioned that one does not wish to limit the scope of this method by assuming a plant with an equal number of inputs and outputs. Rather, at the present time there is no general way of obtaining a square \bar{Q} from a nonsquare \bar{G} . Clearly more work is needed in this area.

When confronted with a nonsquare plant, one does have the option of reverting to the direct Nyquist array method in which analogs results as to dominance, stability and the location of the closed loop transfer function exist [2, pp. 174-179. One usually chooses not to pursue this recourse for several important reasons [2, p. 155].

- 1) Practical experience seems to show that there is a tendency for \hat{G} to be more dominant than \bar{G} .

- 2) When working with inverse systems, the simplicity of relating the open loop system properties to those of the closed loop system adds much additional insight into a design.
- 3) Feedback causes the width of the Ostrowski bands to rapidly become small in the inverse system. That is to say, as feedback is increased in the other loops, we make h_i tend towards \hat{q}_{ii}^{-1} ; the uninversed system has the opposite property and requires all the other loop gains to go to zero.

As our primary interest is in designing for adequate closed loop responses, it is clear why this limitation is accepted and why one is restricted to the inverse system representation.

2.4.1. Matrix Operations

Consider the $(m \times m)$ input compensator matrix $\overline{K(s)}$. For practical reasons $\overline{K(s)}$ should have all its poles and zeros in the open left half plane. This is because it must be physically implemented by a minimum phase, asymptotically stable subsystem. It can be shown [3, p. 209] that such a matrix may be constructed from

$$\overline{K(s)} = \overline{K_a} \overline{K_b(s)} \overline{K_c(s)}. \quad (2.4.1)$$

When working with inverses, this is

$$\hat{K}(s) = \hat{K}_c(s) \hat{K}_b(s) \hat{K}_a \quad (2.4.2)$$

where each of the subscripted square matrices has the following function:

- 1) \hat{K}_a is a permutation matrix, which is to say, in each column and each row all entries are zero except one, which is unity. Its effect is

to transpose the rows of any matrix which it premultiplies. When operating on \hat{G} its function is to renumber the inputs.

- 2) $\hat{K}_b(s)$ is constructed as the product of elementary matrices $\hat{K}_b^{(\ell)}(s)$, each of which has unity determinant and is of the following form:

$$\hat{K}_b^{(\ell)}(s) = \begin{bmatrix} 1 & 0 & \dots & 0 & \dots & 0 & \dots & 0 \\ 0 & 1 & \dots & 0 & \dots & 0 & \dots & 0 \\ \vdots & \vdots & & \vdots & & \vdots & & \vdots \\ 0 & 0 & \dots & 1 & \dots & -z_j^{(\ell)}(s) & \dots & 0 \\ \vdots & \vdots & & \vdots & & \vdots & & \vdots \\ 0 & 0 & \dots & 0 & \dots & 0 & \dots & 1 \end{bmatrix} \quad (2.4.3)$$

where each $z_{ij}^{(\ell)}(s)$ may occur in any off diagonal position and should be a rational function of s , with strictly left half plane poles and zeros. The effect of each $\hat{K}_b^{(\ell)}(s)$ is to subtract from row i a multiple $z_{ij}^{(\ell)}(s)$ of row j from the matrix which premultiplies it. When operating on the matrix $\hat{K}_a \hat{G}(s)$, the effect of the matrix $\hat{K}_b(s)$ is to recombine the parallel paths through the system such that the resulting matrix

$$\hat{K}_b(s) \hat{K}_a \hat{G}(s) = \prod_{\ell=1}^m (\hat{K}_b^{(\ell)}(s)) \hat{K}_a \hat{G}(s) \quad (2.4.4)$$

is diagonally dominant.

- 3) The matrix $\hat{K}_c(s)$ is diagonal and nonsingular. As it represents m single loop controllers, which will be designed from classical frequency loci techniques once sufficient dominance is obtained from (2.4.4), $\hat{K}_c(s)$ should contain strictly left half plane poles and zeros.

In quite an analogous fashion, one may construct the $(m \times m)$ output compensator matrix $\overline{L}(s)$ as

$$\overline{L}(s) = \overline{L}_c(s) \overline{L}_b(s) \overline{L}_a \quad (2.4.5)$$

so that

$$\hat{L}(s) = \hat{L}_a \hat{L}_b(s) \hat{L}_c(s) \quad (2.4.6)$$

where the subscripts indicate matrices whose operations correspond to those in the previous case.

Usually one designs an input compensator with $\hat{K}(s)$ as in (2.4.2) and $\hat{L} = \overline{I}_m$. It is possible to design an output compensator with $\hat{K} = \overline{I}_m$ and \hat{L} as in (2.4.6). This allows some additional freedom in design as the physical situation requires. Very often the matrix $\hat{L}(s)$ will be moved around the loop and actually implemented in the input.

One may always attempt to construct a diagonally dominant \hat{Q} from these elementary row or column operations. Unfortunately, this becomes quite a chore for large systems. Even with a plant of small dimension it is very possible that a satisfactory solution will be overlooked. What is needed is a more systematic method of obtaining dominance.

The first such method which has been tried is to diagonalize $\hat{G}(o)$ with a constant $\hat{K}[6]$. If $\overline{G}(o)$ is nonsingular, this procedure will yield a $\hat{Q}(s)$ which is dominant about $w=0$, but one would be extremely lucky if dominance is maintained throughout the D contour. Similarly, another possibility is to diagonalize $\hat{G}(s)$ as s takes on large imaginary values, but this, too, would just be a guess. Most likely dominance would fail elsewhere on the contour.

One could attempt to combine these two efforts into a more complicated expression, such as

$$\hat{K}(s) = \bar{D}_0 \hat{K}_0 + \hat{K}_1 s \quad (2.4.7)$$

where \hat{K}_0 diagonalizes $\hat{G}(0)$, \hat{K}_1 diagonalizes $\hat{G}(s)$ as s approaches $j\infty$, and \bar{D}_0 is a diagonal matrix. Unfortunately for systems of many dimensions, the necessary inversion of $\hat{K}(s)$ given by (2.4.7) becomes rather complicated, is awkward to calculate exactly, and would be difficult to implement on the actual physical plant. Some of these problems may be overcome if, instead, one used the form

$$\bar{K}(s) = \bar{D}_0 \bar{K}_0 + \bar{D}_1 \bar{K}_1 / s \quad (2.4.8)$$

where \bar{K}_0 diagonalizes $\bar{G}(s)$ as s approaches $j\infty$, \bar{K}_1 diagonalizes $\bar{G}(0)$, and \bar{D}_0 and \bar{D}_1 are each diagonal (gain) matrices.

While either of the forms (2.4.7) or (2.4.8) may be tried, the difficulty is obtaining a dominant \hat{Q} at intermediate frequencies along the Nyquist D contour remains. Furthermore, our interest is still in finding a compensating matrix which is as simple as possible. A satisfactory constant compensating matrix might exist which would suite the purpose. This leads one to investigate possible diagonalization of $\hat{G}(s)$ at $j0 < s < j\infty$. Of course, the dominance of the resulting \hat{Q} will have to be checked on the entire domain of s in question, but a method which gives such a possible candidate is needed.

2.4.2. Pseudodiagonalization [7]

The diagonalization of $\hat{Q}(j\omega)$ by choosing

$$\hat{K} = \overline{G(j\omega)} \quad (2.4.9)$$

would require \hat{K} to be a function of s , as \overline{G} is generally a complex function. One could assume \hat{K} to be constant and formulate the problem as minimizing the sum of the moduli of the off diagonal terms of the resulting $\hat{Q}(s)$. Upon placing some convenient constraint on \hat{K} , it can be shown that the resulting minimization problem is straightforward to formulate, but impossible to solve for a strictly real matrix \hat{K} [7].

This problem may be overcome if one tries instead to minimize the sum of the squares of the moduli of the off diagonal terms. While this may be done at any specific $s = j\omega$, instead the problem is generated at N discrete, independently weighted intervals of frequency ω_r . That is, by considering row j of \hat{Q} , the problem is to find row j of \hat{K} such that the function

$$\sum_{r=1}^N \left\{ \sum_{\substack{k=1 \\ k \neq j}}^m \gamma_r |\hat{q}_{jk}(j\omega_r)|^2 \right\} \quad (2.4.10)$$

is minimized. Here each γ_r is chosen to be positive and real, and represents the weighting factor for each ω_r as $r = 1, 2, \dots, N$. While the following results are based on an input compensator and rows as this is the usual case, analogous results for output compensators or columns may be obtained in a similar manner.

Element $\hat{q}_{jk}(j\omega_r)$ for expression (2.4.10) may be written as

$$\hat{q}_{jk}(j\omega_r) = \sum_{i=1}^m \hat{k}_{ji} \hat{g}_{ik}(j\omega_r) \quad (2.4.11)$$

and while $\hat{g}_{ik}(j\omega_r)$ is complex, it is expressed by real and imaginary parts as

$$\hat{g}_{ik}(j\omega_r) = \alpha_{ik}^{(r)} + j\beta_{ik}^{(r)}. \quad (2.4.12)$$

Then the function (2.4.10) to minimize becomes

$$\sum_{r=1}^N \gamma_r \sum_{\substack{k=1 \\ k \neq j}}^m \left| \sum_{i=1}^m \hat{k}_{ji} (\alpha_{ik}^{(r)} + j\beta_{ik}^{(r)}) \right|^2. \quad (2.4.13)$$

Upon introducing the convenient constraint of

$$\sum_{i=1}^m \hat{k}_{ji}^2 = 1 \quad (2.4.14)$$

and appending the constraint with the Lagrange multiplier λ , expression (2.4.10) requires the minimization of the function

$$\phi_j = \sum_{r=1}^N \gamma_r \left\{ \sum_{\substack{k=1 \\ k \neq j}}^m \left| \sum_{i=1}^m \hat{k}_{ji} (\alpha_{ik}^{(r)} + j\beta_{ik}^{(r)}) \right|^2 + \lambda \left[1 - \sum_{i=1}^m \hat{k}_{ji}^2 \right] \right\}. \quad (2.4.15)$$

Taking the necessary partial derivatives with respect to \hat{k}_{jl} and setting them equal to zero to obtain the required minimum results in

$$\frac{\partial \phi_j}{\partial \hat{k}_{jl}} = 0 = \sum_{r=1}^N \gamma_r \left\{ \sum_{\substack{k=1 \\ k \neq j}}^m \left[2 \left(\sum_{i=1}^m \hat{k}_{ji} \alpha_{ik}^{(r)} \right) \alpha_{lk}^{(r)} + 2 \left(\sum_{i=1}^m \hat{k}_{ji} \beta_{ik}^{(r)} \right) \beta_{lk}^{(r)} \right] - 2\lambda \hat{k}_{jl} \right\}. \quad (2.4.16)$$

For simplicity the row vector \hat{K}_j is defined as

$$\hat{K}_j = (\hat{k}_{jl}). \quad (2.4.17)$$

Similarly the matrix

$$\bar{A}_j = (a_{il}^{(j)}) = \sum_{r=1}^N \gamma_r \left[\sum_{\substack{k=1 \\ k \neq j}}^m (\alpha_{ik}^{(r)} \alpha_{lk}^{(r)} + \beta_{ik}^{(r)} \beta_{lk}^{(r)}) \right] \quad (2.4.18)$$

is introduced. \bar{A}_j is symmetric and real, and at least positive semidefinite by construction, so all of its eigenvalues are real and nonnegative. This

allows (2.4.16) to be written in the form of a standard eigenvalue problem

$$\bar{A}_j \hat{K}_j^T - \lambda \hat{K}_j^T = 0. \quad (2.4.19)$$

While any eigenvector of \bar{A}_j satisfies expression (2.4.19), if one constructs \hat{K}_j from the smallest nonzero eigenvalue of \bar{A}_j , the minimum of the original function (2.4.10) is obtained.

This procedure yields a j th row of \hat{K} which is of unit length and minimizes the sum of the squares of magnitude of the off diagonal terms of row j in $\hat{Q}(j\omega)$. It is clear that this must be done for all the rows $j = 1, 2, \dots, m$. Nothing prevents the use of different values of ω_r or γ_r from (2.4.10), or just a single frequency and unity weighting for all the rows. In any case, the diagonal dominance of the resulting $\hat{Q}(s)$ must always be checked at all points of interest.

The constraint (2.4.14) leads to a relatively easy computational problem. Unfortunately, while satisfying (2.4.14) it often will simultaneously lead to unacceptably small values of \hat{q}_{jj} . To overcome this, the alternate constraint

$$\sum_{r=1}^N |\hat{q}_{jj}(j\omega_r)| = 1 \quad (2.4.20)$$

may be substituted for (2.4.14). A similar analysis shows that the more difficult solution of the eigenvector problem

$$\bar{A}_j \hat{K}_j^T - \lambda \bar{E}_j \hat{K}_j^T = 0 \quad (2.4.21)$$

is required. Here \bar{A}_j is as in (2.4.18), and \bar{E}_j is the symmetric, positive semidefinite matrix defined by

$$\bar{E}_j = (e_{jl}^{(j)}) \sum_{r=1}^N \gamma_r [\alpha_{ij}^{(r)} \alpha_{lj}^{(r)} + \beta_{ij}^{(r)} \beta_{lj}^{(r)}]. \quad (2.4.22)$$

The idea of pseudodiagonalization may be extended by considering compensating matrices dependent upon frequency as in (2.4.7) where higher power terms in s may be included as needed. The general comments of the previous subsection still motivate one against such a choice. The problem is further compounded in that minimization techniques as this in no way guarantees a resulting matrix $\overline{K(s)}$ which has strictly left half plane poles and zeros.

As can be seen, obtaining open loop dominance is very often a function of the physical situation at hand. The degree of dominance is usually dependent upon the extent of complexity the system designer will allow to be actually implemented. Certainly before these methods can be put into a more general use, additional work on obtaining dominance is needed.

2.4.3. Closed Loop Dominance via Dynamic $\overline{F(s)}$

One may consider obtaining dominance in the closed loop system. If $\overline{G(o)}$ is nonsingular, there always exists a constant input compensating matrix \bar{K} as

$$\bar{K} = \hat{G}(o) \quad (2.4.23)$$

because $\overline{G(o)}$ is real. This will make $\hat{Q}(o)$ diagonal and dominance will remain for at least a small region close to the origin in the complex plane. Consider

$$\overline{F(s)} = \overline{F(o)} + \overline{F_1(s)} \quad (2.4.24)$$

where $\overline{F(o)}$ is diagonal and scalar. $\overline{F_1(s)}$ has zero diagonal elements. Its off diagonal elements, which are a function of frequency, are chosen so that the resulting dominance of \hat{H} in the closed loop system is improved. As $F_1(o)$ is constructed to be zero, the resulting $F(s)$ yields transient compensation of the interaction.

The scope of the method has not yet been well defined [8]. Such a method has never actually been physically implemented. If it would be, great care in the amount of phase advance introduced and its effect on the dynamic behavior of the system should be carefully considered. The sensitivity of the system with respect to changes in the plant will also have to be investigated, as a small change in \overline{Q} may severely change the degree of dominance when cancellation of terms in $\overline{F} + \hat{Q}$ are large.

2.5. Stability

In multivariable control problems, the allowable gain in one loop is very often a complicated function of the gains in the other loops. For design purposes one would like to have a good idea of what combinations of gains result in asymptotic closed loop behavior. Permissible feedback values could then be chosen, perhaps with the additional requisite that stability is maintained as these gains vary a prescribed amount from their design values. In addition, it may be desirable to design the closed loop system so that it does not become unstable in the case of some catastrophic event,

such as a transducer or actuator failure. Before showing how this is possible some results of a general nature are necessary.

2.5.1. Frequency Domain Criteria for Stability

Theorem 3 [2, p. 141]: Let \bar{Q} have an inverse \hat{Q} as defined by (2.2.3), which has p_o poles in the right half plane. Let $|\hat{Q}|$ map the Nyquist contour D into $\hat{\Gamma}_Q$ and $|\hat{H}|$ map D into $\hat{\Gamma}_H$. As s goes once around D , $\hat{\Gamma}_Q$ will encircle the origin \hat{N}_Q times, while $\hat{\Gamma}_H$ encircles the origin \hat{N}_H times, where all positive encirclements are clockwise. Then the resulting closed loop system is asymptotically stable if and only if

$$\hat{N}_Q - \hat{N}_H = p_o. \quad (2.5.1)$$

The proof of this theorem requires showing that the return difference matrix relates the open loop poles to the closed loop poles and must be motivated on a very general ground from the theory of polynomial system matrices and the "Principle of the Argument" in complex number theory. As it is not the exact result we are seeking for design purposes, its proof may be found elsewhere [2, pp. 131-141].

Theorem 4 [2, p. 143] (Dominant Encirclement Theorem): Let the $(k \times k)$ rational complex matrix $\overline{Z(s)}$ be dominant on any simple closed contour C which has on it no pole of $z_{ii}(s)$ for all $i=1,2,\dots,k$. Let $z_{ii}(s)$ map C into Γ_i , which encircles the origin N_i times, and $|Z(s)|$ map C into Γ_z , which encircles the origin N_z times. All clockwise encirclements are again taken to be positive. Then

$$N_z = \sum_{i=1}^k N_i. \quad (2.5.2)$$

The proof of the Dominant Encirclement Theorem relies on Gershgorin's Theorem and a necessary corollary. It is rather lengthy and may be found elsewhere [2, pp. 25-26].

From these two theorem statements, a number of different stability theorems may be motivated depending upon which matrices are assumed to be dominant, and whether or not one is working with inverse systems. Rather than stating all of the possibilities, one chooses instead to give a result which is the most useful one for design purposes.

Theorem 5: (Generalized Graphical Inverse Nyquist Stability Criterion)*

Let \hat{Q} and \hat{H} be dominant on D , and assume that \bar{F} is diagonal and independent of s .

Suppose each of the Gershgorin bands based on the diagonal elements \hat{q}_{ii} of \hat{Q} exclude the origin and the critical point $(-f_i, 0)$. Let these bands encircle the origin \hat{N}_{qi} times and encircle the critical point $(-f_i, 0)$ \hat{N}_{hi} times. Then the resulting closed loop system is asymptotically stable if and only if

$$\sum_{i=1}^k \hat{N}_{qi} - \sum_{i=1}^k \hat{N}_{hi} = p_o \quad (2.5.3)$$

where p_o is the same as in Theorem 3.

Proof: When \bar{F} is diagonal and independent of s , exclusion of the origin by a set of bands based upon $\hat{h}_{ii} = f_i + \hat{q}_{ii}$ is the same as the exclusion of the point $(-f_i, 0)$ by a set of bands based on \hat{q}_{ii} . While noting that the premise of the Dominant Encirclement Theorem applies and gives the total number of

* Additional theorems and a more thorough development leading up to the statement of this theorem may be found in Rosenbrock [2, pp. 143-154].

required encirclements in both the open and closed loop system, Theorem 3 is used and the required results are obtained.

This theorem represents the Generalized Inverse Nyquist Stability Criterion for multivariable systems and has the graphical design implications as stated. It should be mentioned that it is possible to exclude the origin with one set of bands, for example based on rows, while excluding the critical point $(-f_1, 0)$ by a set of bands based on columns, as is otherwise implicitly obvious in the use of the word dominance. While this theorem is of most use in its present form, it sometimes is possible to make \hat{H} dominant but not \hat{Q} . In this special case, there is the following result

Corollary 1 [2, p. 145]: Let \hat{H} be dominant on D and satisfy the dominant encirclement theorem. Then the closed loop system is asymptotically stable if and only if

$$\sum_{i=1}^k \hat{N}_{hi} = p_Q - z_Q - p_o \quad (2.5.4)$$

where \hat{N}_{hi} and p_o are as in Theorem 5 and p_Q and z_Q are the poles and zeros of $|\bar{Q}|$.

Sketch of Proof: (2.5.1) in Theorem 3 states that for asymptotic stability, one must have

$$\hat{N}_Q - \hat{N}_H = p_o \quad (2.5.5)$$

while the dominant encirclement theorem states

$$\hat{N}_H = \sum_{i=1}^k \hat{N}_{hi}. \quad (2.5.6)$$

An application of the Principle of the Argument gives

$$\hat{N}_Q = p_Q - z_C \quad (2.5.7)$$

and combining (2.5.5) through (2.5.7) yields the desired statement. Notice also that these results can be useful if $|\bar{Q}|$ is known algebraically.

Underlying the stability theorems in either the open or closed loop case is the idea of diagonal dominance. If a system cannot be made dominant, the theorems say nothing.

2.6. A Finishing Remark

If any design method is to be of good use it must be physically practical, both in terms of implementation and measurement. In particular, the inverse Nyquist array method fails in this respect as those values of complex frequency which correspond to the semicircular arc on the Nyquist contour D have no physical significance.

There is a natural desire to be relieved from considering this portion of the Nyquist contour in a design. Not only is it impossible to perform actual measurements on the system for these values of s , dominance may fail in this region as well.

While theorems have been presented which lead one to believe this is possible, after Rosenbrock, the following is stated as an assumption [2, pp. 153-155]:

'Assumption: Those poles which determine stability cross the segment of the imaginary axis between $s = -j\omega_0$ and $s = j\omega_0$, where ω_0 is some specified (real) frequency.'

Therefore, one can investigate stability by considering imaginary values of s from $0 \leq s \leq j\omega_1$. Here ω_1 is greater than ω_0 and $\omega_1 \geq R$, as defined by

the inclusion of all right half plane poles and zeros of the transfer functions by the Nyquist contour D.

Being able to exclude this semicircular arc makes the practical considerations of the design method complete.

CHAPTER 3

COMPUTER PROGRAMMING

3.1. Introduction

Rosenbrock's method, as outlined in the previous chapter, represents a feasible alternative in the design of multivariable control systems. It does require a rather large amount of calculation, even for plants of low order. This could be done by hand and the results graphed accordingly, but the necessary time and labor would make such an approach extremely unappealing.

Fortunately, with the present availability of high speed, time sharing computers this computational effort can be completely eliminated. Furthermore, with the advent of graphical display terminals, these results may be immediately graphed and examined. Finally, with the development of interactive software, a suitable control scheme can be built up as insight gained into the system is quickly reapplied until an acceptable design is reached.

Such an effort has been undertaken using a Tektronix graphics terminal with supporting software on the Coordinated Science Laboratory's DEC-10 computer.

Programming in the form of twenty-one subroutines has been developed in DEC system-10 FORTRAN code [9] and may be compiled using either the F10 or the F40 FORTRAN compiler. Graphic routines have been written with the aid of the Tektronix Advanced Graphics AG-II supportive software package [10].

All the subroutines are functionally oriented and have been kept general, to be useful in the design of any control problem. The routines operate on data which is stored in a common block.

As conversational software for the analysis and design of systems has become in vogue [11], it is hoped that these subroutines will be included as an option package in an interactive software program. At that time the common data base will have to be appropriately modified to conform with the conventions already established in the particular program used.

It is possible to combine the written subroutines themselves into an interactive software program. An example of how this may be done is included with the next chapter in investigating a particular design example.

3.2. Common Block Usage

The common data area has been divided into nine separately labeled segments, each of which serves its own functional importance. For efficient computer utilization when solving a specific problem, these common areas should be dimensioned as the particular situation dictates. The routines are therefore stored in separate files without the common data base and commented as to which labeled common block area is required.

COMMON /A1/ Block

Labeled common area /A1/ stores all of the system matrices as shown in the block diagram structure included in Figure 2.1.2. Matrices

$\overline{K(s)}$ and $\overline{L(s)}$ are each broken into constant matrices, XKA and XLA, and into frequency dependent matrices, XKB and XLB, respectively.

The transfer function of the plant is stored in the four dimensional array G as follows: considering location

$$G(I,J,K,L) \quad (3.2.1)$$

defines a storage position where

I = row index of transfer function

J = column index of transfer function

K = coefficient of (K-1) power of s in G(I,J)

L = 1 for the numerator polynomial

= 2 for the denominator polynomial.

Then when dimensioning G, it is clear that I and J must be equal and correspond to the number of inputs or outputs, K must be one more than the highest power of s occurring in any entry of G, and L is 2.

The coefficients of the polynomial terms in matrices XKB and XLB are referred to in exactly the same manner and are dimensioned as needed.

Matrices XKA, XLA, and F are all square, constant, and of appropriate dimension.

COMMON /A2/ Block

Labeled common area /A2/ stores seven single variables. In order of occurrence they are:

OMEGØ - real variable denoting the starting frequency of the Nyquist path.

It will almost always be defined to be zero, but is included so that a region on the jw axis not including the origin may be investigated.

OMEGND - real variable corresponding to the end frequency on the Nyquist path on the $j\omega$ axis. Referring to Figure 2.1.1, it is also the length of the radius R defining the semicircular arc in the right half plane.

DTOMEG - real variable corresponding to the change in frequency from OMEG0 to OMEGND as one travels up the $j\omega$ axis.

ISTART - integer variable denoting the starting location in arrays yet to be defined. While it usually is 1, it has been included so we may plot any group of points corresponding to a particular segment of the frequency loci.

IEND - integer variable; it defines an end location as ISTART defines a starting location. On the $j\omega$ axis, one is considering

$$IEND = [(OMEGND - OMEG0) / DTOMEG] + 1 \quad (3.2.2)$$

total number of such points.

NP - integer variable denoting the number of points between each Gershgorin or Ostrowski band plotted.

MO - integer variable correspond to number of inputs or outputs. In reference to Chapter 2, it is the order of the system or k.

COMMON /A3/ Block

Labeled common area /A3/ is a scratch area used as temporary storage in many of the computational routines. It has been further divided into three separate areas.

Area /A31/. Stores two square complex matrices C1 and C2, each which should be dimensioned (MO x MO). It also has two real variable vectors S1 and S2 of length MO.

Area /A32/. Stores two real variables, three dimensional arrays XR1 and XI1, which correspond to real and imaginary parts of the Nyquist map. Each should be dimensional as (MOX MOX IEND). When referring to the point in the complex plane stored in position

$$(XR1(I,J,N),XI1(I,J,N)) \quad (3.2.3)$$

then one is considering the (I,J) input-output relationship at a frequency corresponding to the Nth point on the frequency loci. For points on the $j\omega$ axis, the frequency is

$$\omega = (N-1) * DTOMEG \quad (3.2.4)$$

as N will take on values from ISTART to IEND.

COMMON A/33/. Stores two real variables, three dimensional arrays XR2 and XI2, each dimensioned as (MOX MOX IEND). They define positions and serve a function similar to arrays XR1 and XI1 in area /A32/.

COMMON /A4/ Block

Labeled common area /A4/ stores the real and imaginary parts of the open loop inverse transfer function \hat{Q} . It does so in two, three dimensional real variable arrays, QR and QI respectively. Each is dimensioned as (MOX MOX IEND) and the (I,J,N) storage position contains information in the same manner as defined around (3.2.3) and (3.2.4).

COMMON /A5/ Block

Labeled common area /A5/ stores information including the value of the Gershgorin bands in the four dimensional real value array BAND. Stored in location

BAND (I,J,N,L)

is the sum of the magnitudes not including the (I,J) magnitude at frequency N of the Q values from area /A4/. The sum of the row magnitudes are stored in L=1 and the sum of the column magnitudes are stored in L=2. Array BAND should be dimensioned as BAND (MO,MO,IEND,2).

COMMON /A6/ Block

Labeled common area /A6/ contains one three dimensional real valued array OST, which stores the values of the Ostrowski bands. In location OST (I,N,L) is the Ostrowski band based on \hat{h}_{II} at frequency N where again L=1 for the band based on rows and L=2 for the band based on columns. Array OST should therefore be dimensioned as OST (MO,IEND,2).

COMMON /A7/ Block

Labeled common area /A7/ contains two important real value arrays XS and XC. XS is square and dimensioned as XS(MO,MO). XC is four dimensional and dimensioned the same as G in area /A1/ as defined by (3.2.1). In the subroutines which follow, calculations are performed with matrices XS and XC which will take on values from area /A1/ and thus allow the design of either input or output compensators.

3.3. Subroutines

The subroutines have been divided into three separate sections which in some respect correspond to the operation which they perform. Each individual routine is stored separately on disk under a file name which

corresponds to the subroutine name with an extension of SAV. Common block usage other than COMMON /A2/ has been appropriately commented but not included. Subroutines have further been commented as to what interactive procedures they perform, if any, and what additional subroutines are called in that routine, if any. A listing of the subroutines and a brief description of each of their functions is included in Appendix 2.

3.3.1. Library COMP

Library COMP contains seven computational subroutines useful in performing the necessary calculations at various stages of a design.

Subroutine MAP

The subroutine MAP computes the complex value of the transfer function matrix $\overline{G(s)}$ as s takes on values around the Nyquist path as defined by the real variables in COMMON /A2/.

The calling sequence is

```
CALL    MAP (ICIRCL)
```

where

ICIRCL = 0 maps the value of $\overline{G(s)}$ for strictly imaginary values of s on the $j\omega$ axis from OMEG0 to OMEGND in discrete steps of DTOMEG.

The total number of these points is

$$NTIMES = ((OMEGND-OMEG0)/DTOMEG) + 1. \quad (3.3.1)$$

ICIRCL = 1 maps the value of $\overline{G(s)}$ as s goes around the semicircle of radius OMEGND. It does this in 99 discrete steps.

ICIRCL = 2 (or any other integer) causes s to accept values along the entire Nyquist path. It will do so in

$$(NTIMES + 99 + NTIMES) \quad (3.3.2)$$

discrete steps.

The resulting value of $\overline{G(s)}$ is a complex number whose real part is stored in XR1 and whose imaginary part is stored in XI1 in COMMON /A32/, for each g_{ij} and the N such discrete steps performed.

If a pole of G is encountered anywhere on the imaginary axis, subroutine MAP types on the screen the message

A POLE AT OMEGA = XXXXX.XX FOR $G(X,X,XXX)$.

It then adjusts the value of the complex frequency variable s accordingly to

$$s = (\text{OMEGA}, -\text{DTOMEG}) \quad (3.3.3)$$

so that these poles are included inside the contour D .

Subroutine MAG

The subroutine MAG takes the complex value of the functions implicitly defined in array XC and stores them in the labeled common area /A32/ in arrays XR1 and XI1. It does so at discrete frequencies defined by (3.2.4) where

$$N = 1, 2, \dots, \text{IEND}. \quad (3.3.4)$$

This subroutine is used in the design of frequency dependent compensators. The appropriate array from COMMON /A1/ is loaded into XC and MAG computes its complex value. Note that if one were to set $XC = G$, then MAG performs the same function as calling MAP(\emptyset).

Singularities of functions in XC have been programmed around in the same fashion as in routine MAP. The user is notified of these as the statement

A POLE AT OMEGA = XXXXX.XX FOR FUNCTION (X,X,XXX)

is typed on the terminal screen.

The calling sequence is

CALL MAG

and has no calling arguments.

Subroutine MIMP

The subroutine MIMP performs the matrix multiplication of constant and matrix XS from COMMON /A7/ with the complex matrix represented by (XR1,XI1) from COMMON /A32/. It does so for the N positions defined by (3.3.4). The resulting complex matrix is appropriately stored in labeled common /A4/ area by its real part in array QR and its imaginary part in array QI.

Routine MIMP finds its usefulness in the design of constant matrix compensators. It also presents an easy means of transferring data from labeled common area /A32/ to labeled common area /A4/ when matrix XS is set equal to the identity matrix \tilde{I}_m . As all the graphics routines plot frequency loci from data contained in arrays QR and QI from common /A4/, this is a necessary transfer to perform in persuing the map of $\overline{G(s)}$.

The calling sequence is

CALL MIMP

where no calling arguments are needed.

Subroutine MCMP

The subroutine MCMP is similar to routine MIMP except that it performs the multiplication of two complex matrices. The routine takes the product of the complex matrix defined by XR1 and XI1 from COMMON /A32/ with the complex matrix defined by XR2 and XI2 from COMMON /A33/. It then stores the resulting matrix, again by real and imaginary parts in QR and QI in COMMON /A4/. This is performed for all N as in (3.3.4).

The routine is useful in the design of frequency dependent compensators. The calling sequence, which has no arguments, is

CALL MCMP

Subroutine GERSH

The subroutine GERSH computes the moduli of the Q values from COMMON /A4/ and stores the resulting real number in array BAND in COMMON /A5/ as

$$\text{BAND}(I, J, N, 1) = \sum_{\substack{k=1 \\ k \neq J}}^{M0} |(\text{QR}(I, k, N), \text{QI}(I, k, N))| \quad (3.3.5)$$

$$\text{BAND}(I, J, N, 2) = \sum_{\substack{k=1 \\ k \neq I}}^{M0} |(\text{QR}(k, J, N), \text{QI}(k, J, N))|. \quad (3.3.6)$$

When $I = J$ the value stored in $\text{BAND}(I, J, N, 1)$ represents the radius d_i from expression (2.2.3) for Gershgorin bands based on the row \hat{s} of \hat{Q} , while the value stored in $\text{BAND}(I, J, N, 2)$ represents the radius d_i' from expression (2.3.4) for Gershgorin bands based on the columns of \hat{Q} . This is for all N as in (3.3.4). Note that those values in BAND for $I \neq J$ are not actually Gershgorin bands but the appropriate sum of the magnitudes including

the diagonal term. This information may be useful in determining how a preliminary input compensator might be constructed in order that the diagonal dominance is achieved, and therefore has been included.

Upon executing the statement

CALL GERSH

these magnitudes are computed from QR and QI and stored in BAND.

Subroutine OSTROW

The subroutine OSTROW computes the Ostrowski bands and places the resulting value of the circular radius in array OST as discussed in the labeled common area /A6/ section. These computed values, as defined by (2.3.11) and (2.3.12), are based on the \hat{Q} values from COMMON /A4/, their resulting diagonal BAND values from COMMON /A5/ and the current feedback gains in diagonal matrix XS from COMMON /A7/. In the closed loop inversed system, as one changes the amount of feedback, the size of the Ostrowski bands change. It then becomes necessary only to call subroutine OSTROW for each change of feedback, as the \hat{Q} values and BAND values remain constant as the loops are closed.

The Ostrowski bands are calculated by executing the statement

CALL OSTROW.

Subroutine MMIJ

The subroutine MMIJ determines the minimum and maximum points on the composite plot of a Nyquist diagram with its Gershgorin bands. It therefore requires use of the \hat{Q} values in COMMON /A4/ and the BAND values in COMMON /A5/.

The calling sequence is

```
CALL MMIJ(I,J,XMIN,XMAX,YMIN,YMAX)
```

where

I = row index of composite plot

J = column index of composite plot

XMIN = minimum horizontal value of composite plot

XMAX = maximum horizontal value of composite plot

YMIN = minimum vertical value of composite plot

YMAX = maximum vertical value of composite plot.

Subroutine MMIJ searches the composite diagram for $I = J$ and compares the MIN and MAX variables with the outer points of the composite diagram. For $I \neq J$ the same search is performed on the frequency loci without the Gershgorin bands. In either case it is necessary to initialize the MIN variables to a large positive value and the MAX values to a large negative value in calling the routine.

This subroutine serves two purposes when used in conjunction with the graphics routines. First, it determines where the screen margins should be set to when drawing a composite diagram, as the supportive software for the graphics terminal has no way of doing this when the Gershgorin bands are included. Second, it provides a means by which a common scale may be determined when plotting several such diagrams on the screen at once.

3.3.2. Library GRAPH

Library GRAPH contains six subroutines which display all the information necessary in performing a design.

Subroutine PTNY1

The subroutine PTNY1 plots the \hat{Q} values in COMMON /A4/ from position ISTART to position IEND on the terminal screen. The supportive software automatically sets screen limits by the CALL CHECK statement if these have not been previously set elsewhere. The CALL DISPLAY statement causes the Nyquist plot and axis to be graphed on the screen.

The calling sequence is

CALL PTNY1(I,J)

where

I = row index of loci plotted.

J = column index of loci plotted.

Subroutine PTGR1

The subroutine PTGR1 causes the Gershgorin bands to be graphed over the Nyquist plot by use of the supportive software routine CPLOT. It plots these bands starting from position ISTART, and plots one band every NP positions, to position IEND.

The calling sequence is

CALL PTGR1(I,J,IRC)

where

I = row index of bands plotted

J = column index of bands plotted

IRC = 1 for Gershgorin bands based on rows

= 2 for Gershgorin bands based on columns.

Subroutine PTOST1

The subroutine PTOST1 plots the Ostrowski bands in exactly an analogous fashion as routine PTGR1 plots the Gershgorin bands.

The calling sequence is

CALL PTOST1(I,IRC)

where

I = diagonal position of h_{ii}

IRC = 1 for Ostrowski bands based on rows

= 2 for Ostrowski bands based on columns.

Subroutine PTBIG1

The subroutine PTBIG1 interactively plots a single Nyquist plot and the corresponding Ostrowski bands using the maximum possible terminal screen size.

Upon executing the calling statement

CALL PTBIG1

the graphic display panel is erased and the message

TYPE \emptyset TO RETURN, OR TYPE I,1/2 FOR H(I,I) AND OST ROWS/COLUMNS

is typed at the top of the screen. Entering \emptyset again erases the screen and returns execution to the calling program. If values are entered, the screen is cleared and the supportive software is adjusted accordingly so that the desired composite diagram is plotted on the screen using routines PTNY1 and PTOST1. The same message as before is then typed at the very upper edge of the screen, allowing the process to repeat itself as often as needed.

Subroutine MAK4

The subroutine MAK4 causes the four Nyquist diagrams corresponding to the upper left (2X2) square portion of \hat{Q} to be graphed on the screen. It also plots the Gershgorin bands for the two diagonal entries of the defined submatrix if desired. While the system under investigation could be constructed so that this routine would display something useful, its most meaningful application is for a two input, two output system.

The routine first searches all four of the composite diagrams for the maximum and minimum horizontal and vertical points, by four successive calls to subroutine MMIJ. This data is then used with some supportive software routines to set the scale of each plot to be equal. The Nyquist diagrams are graphed on the screen with subroutine PTNY1 and the Gershgorin bands, if desired are plotted on the diagonal loci with routine PTGR1.

The calling sequence is

CALL MAK4 (IRC)

where

IRC = 0 causes no Gershgorin bands to be plotted

= 1 causes Gershgorin bands based on the row elements in \hat{Q} to be plotted

= 2 causes the Gershgorin based on the column elements in \hat{Q} to be plotted.

Subroutine DRW

Subroutine DRW is the major interactive graphics routine. It basically serves a routing function and relies on the previous subroutines to actually perform a given operation. As such, it does not deal with any

physical data and does not require any common block usage. Included within subroutine DRW is entry point DRW1. A flow chart of the routine is included as Figure 3.3.1.

Upon executing the statement

CALL DRW

the message

PLOT DIAGRAM: TYPE: \emptyset FOR NO, 1 FOR YES

is typed on the terminal screen. Entering \emptyset causes execution to return to the calling program, while entering 1 causes execution to proceed to entry point DRW1. This entry point has been included as usually one determines whether or not a graphical output is desired from the calling program. If this is the case and it has been previously determined that a plot is desired one may instead execute the statement

CALL DRW1 .

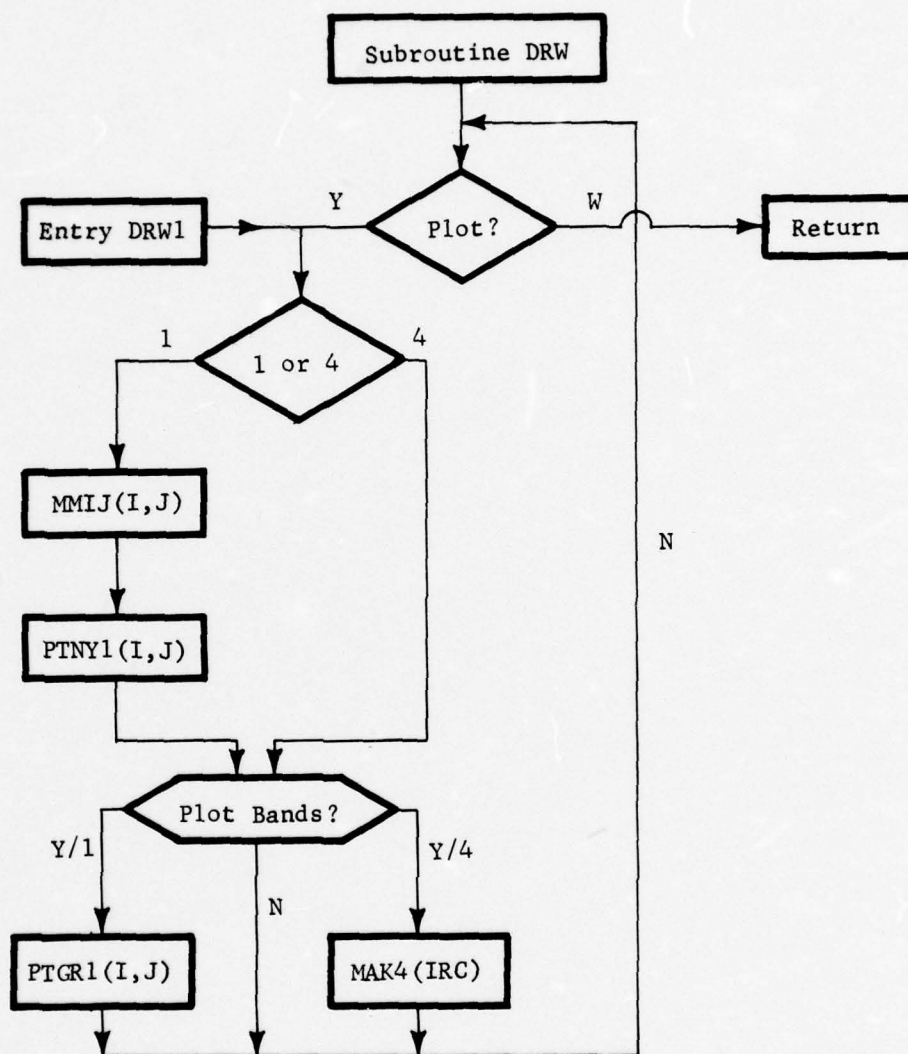
In either case, at this point the message

ENTER -0,0- FOR 4 or -I,J- for Q(I,J)

is typed on the terminal screen. Entering nonzero integer values causes subroutine MMIJ to search for the necessary MAX and MIN values, the scaling is set and subroutine PTNY1 plots the appropriate frequency loci.

If two zeros are entered, or, after the single loci is graphed, the message

PLOT BANDS? TYPE \emptyset FOR NO, 1 FOR ROWS, 2 FOR COLUMNS



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Figure 3.3.1. Flow chart for subroutine DRW.

is typed on the terminal screen. In the case of plotting four diagrams, subroutine MAK4 with the proper band identifying argument is called and the four frequency loci with the desired bands are graphed on the terminal screen. In the case of a single plot, if Gershgorin bands are to be plotted, subroutine PTGR1 with the proper band identifier is called.

In either case the routine then routes itself back to its beginning and the original message

PLOT DIAGRAMS? TYPE 0 FOR NO, 1 FOR YES

is typed on the screen as before, allowing the process to repeat itself as many times as desired.

3.3.3. Library AUX

Library AUX contains eight subroutines of an auxiliary nature which perform the interactive modification, disk storage or outputting of various variables.

Subroutine MODIA2

The subroutine MODIA2 allows the three integer variables ISTART, IEND, and NP in labeled common area /A2/ to be modified from the computer console during a design.

Upon executing the calling statement

CALL MODIA2

the message

ENTER ISTART, IEND, NP, OR -, - AND RETURN

is typed on the screen. Entering three integer values appropriately modifies these variables, while a comma for any entry causes them to remain

the same. The routine then types the three values on the screen in a unformatted output so there is no mistake as to what values they have become.

Subroutine MODXS

The subroutine MODXS performs the interactive modification or output of matrix XS in COMMON /A7/ as diagramed in Figure 3.3.2.

Upon executing the calling statement

CALL MODXS

the message

MODIFY XS? TYPE 0 FOR NO, 1 FOR YES

is typed on the screen. Entering 1 causes the statement

TYPE I,J,X IN XS(I,J) = X

to be typed on the screen. After the values are entered, the first message

MODIFY XS? TYPE 0 FOR NO, 1 FOR YES

again appears, allowing additional entries in XS to be changed.

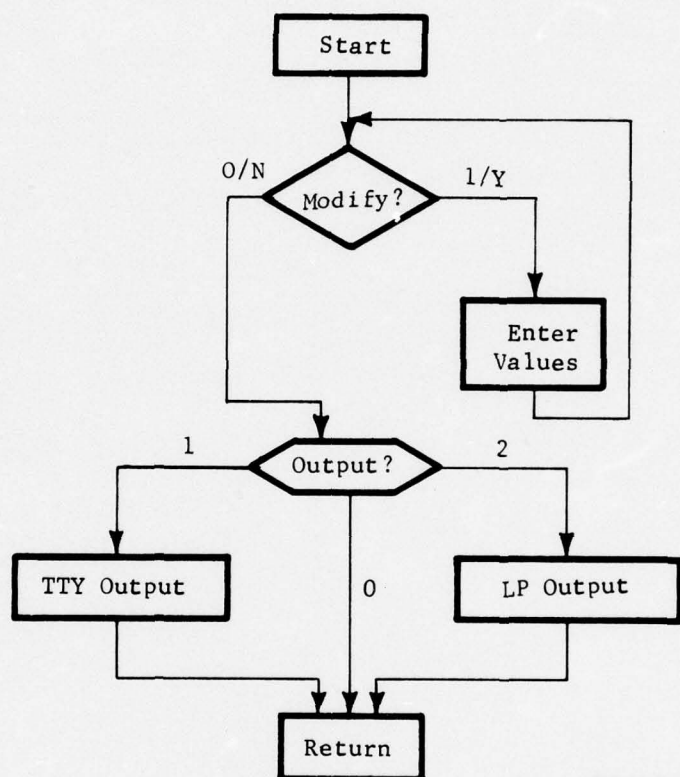
Upon returning 0 the statement

OUTPUT XS? TYPE: 0 FOR NO, 1 FOR TTY, 2 FOR LP

is typed on the screen. After entering the appropriate value the output is performed and execution is returned to the calling program.

Subroutine MODXC

The subroutine MODXC performs the interactive modification and output of array XC as routine MODXS does for matrix XS. This again may be represented by the diagram included as Figure 3.3.2.



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Figure 3.3.2. Functional flow diagram for subroutines MODXS and MODXC.

The calling statement is

```
CALL MODXC .
```

This causes the statement

```
MODIFY XC? TYPE  $\emptyset$  FOR NO, 1 FOR YES
```

to be typed on the terminal screen. When returning 1 the statement

```
I,J,N,X,1/2 IN XC(I,J) = X*(S)**N,NUM/DOM
```

is typed on the screen and after the values are entered, the original message again appears, allowing possible further modification in array XC.

Entering \emptyset causes the statement

```
OUTPUT XC? TYPE  $\emptyset$  FOR NO, 1 FOR TTY, 2 FOR LP
```

to appear on the screen. Returning a value causes the appropriate output to be performed and execution is then returned to the calling program.

Subroutine QOUT

The subroutine QOUT causes the \hat{Q} values in COMMON /A4/ from positions ISTART to IEND to be dumped for numerical investigation.

The calling sequence is

```
CALL QOUT(IDUM)
```

where

IDUM = 1 for terminal screen output

IDUM = 2 for line printer output

Subroutine QDSKWR

The subroutine QDSKWR stores all the \hat{Q} values in COMMON /A4/ on disk file Q1.DAT.

Upon executing the calling statement

CALL QDSKWR

the message

STORE Q VALUES? TYPE \emptyset FOR NO, 1 FOR YES

is typed on the terminal screen. Returning \emptyset causes a return to the calling program. Typing 1 causes the formatted storage of the integer 1 followed by

I,J,N,QR(I,J,N),QI(I,J,N)

on the disk file Q1.DAT. These values may be perused after execution by outputting the file, or used for the next program run by using the subroutine QDSKRD.

Subroutine QDSKRD

The subroutine QDSKRD reads the Q values from disk. It assumes that the disk file Q1.DAT exists and looks at the value stored in the second space of its first line. If this value is \emptyset , a return is made to the calling program and the \hat{Q} values are not read. If instead this value is 1, the \hat{Q} values are read and stored in arrays QR and QI in COMMON /A4/. The routine also adjusts the variable IEND in COMMON /A2/ as the largest Nth position read into these arrays.

The calling sequence is

CALL QDSKRD(IDUM)

where

IDUM = \emptyset on return indicating Q values were not read

= 1 on return indicating Q values were read and stored.

Subroutine OUT

The subroutine OUT outputs all the system matrices stored in COMMON /A1/ and all the programming variables from COMMON /A2/ to either the terminal screen or the line printer. It also has the ability to call subroutine QOUT for a look at the current Q values from ISTART to IEND.

Upon executing the calling statement

CALL OUT

the message

FOR OUTPUT TYPE

Ø FOR NONE, 1 FOR TTY, 2 FOR LP

appears on the terminal screen. Typing Ø causes a return to the calling program. Typing 1 or 2 causes the message

FOR SYSTEM OUTPUT TYPE Ø

FOR Q OUTPUT TYPE 1.

Entering Ø causes the system and programming variables to be dumped, whereas typing a 1 causes the subroutine QOUT to be called.

In either case, after the appropriate output is performed, the message

TYPE Ø TO CONTINUE, TYPE 1, TO DO AGAIN

to appear on the terminal screen. Entering Ø causes execution to return to the calling program, while entering 1 causes the original message

FOR OUTPUT TYPE

Ø FOR NONE, 1 FOR TTY, 2 FOR LP

to be typed on the screen. This allows the process to repeat itself again, if desired.

Subroutine FINISH

The subroutine FINISH is executed by the calling statement

CALL FINISH.

The message

YOU ARE DONE: BEST PRINTOUT SYSTEM!!!

is typed on the terminal screen and subroutine OUT is called so that the current system values may be saved. The screen is then erased and subroutine QDSKWR is called so that Q values may be stored for further use. Execution is then terminated.

4. DESIGN ORIENTATION AND EXAMPLE

4.1. Introduction

The subroutines introduced in the previous chapter have been combined into a main program and three design packages. These have been applied to the particular design problem of a lateral autopilot to maintain heading and roll attitude.

The programming presented here is oriented to the two input, two output autopilot plant using input compensators. Output compensators may be designed in an analogous fashion by modifying the 'work' matrices from COMMON /A7/. Particular attention should be given to the order of matrix multiplication for these types of designs. Plants of higher dimensions cause no problem either, if the common data area is appropriately constructed as was discussed in Section 3.2.

4.2. Autopilot Plant

The lateral motions of an aircraft as defined by the roll, yaw and sideslip angles are coupled together, but are almost completely uncoupled to the longitudinal or pitch and plunge motions. The resulting perturbation equations to first order of the fifth order system describing these lateral motions are found [12] to be:

$$\dot{\beta} + r = \frac{Y_{\beta}}{mV} \beta + \phi \quad (4.2.1)$$

$$\dot{r} + \frac{I_{xz}}{I_{zz}} \dot{p} = \frac{N_{\beta}}{I_{zz}} \beta + \frac{N_r}{I_{zz}} r + \frac{N_p}{I_{zz}} p + \frac{N_{\delta r}}{I_{zz}} \delta r \quad (4.2.2)$$

$$\dot{p} + \frac{I_{xz}}{I_{xx}} \dot{r} = \frac{l_{\beta}}{I_{xx}} \beta + \frac{l_r}{I_{xx}} r + \frac{l_p}{I_{xx}} p + \frac{l_{\delta a}}{I_{xx}} \delta a \quad (4.2.3)$$

$$\dot{\phi} = p \quad (4.2.4)$$

$$\dot{\psi} = r \quad (4.2.5)$$

State variables are defined as

ϕ = roll angle

p = roll angular velocity

ψ = yaw angle

r = yaw angular velocity

β = sideslip angle

and inputs are

δr = rudder deflection

δa = aileron deflection

where (4.2.2) and (4.2.3) must be solved simultaneously for a proper state space description. Doing so and using typical values of the coefficients for an aircraft weighing 100,000 pounds flying at 500 miles per hour at 30,000 feet [12], the resulting state space representation of the plant is

$$\begin{bmatrix} \dot{\beta} \\ \dot{r} \\ \dot{p} \\ \dot{\phi} \\ \dot{\psi} \end{bmatrix} = \begin{bmatrix} -.0297 & -1.0 & 0 & .0438 & 0 \\ .331 & -.0042 & -.046 & 0 & 0 \\ -1.135 & .238 & -.795 & 0 & 0 \\ 0 & 0 & 1.0 & 0 & 0 \\ 0 & 1.0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \beta \\ r \\ p \\ \phi \\ \psi \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ -.38 & .067 \\ -.04 & 1.587 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \delta r \\ \delta a \end{bmatrix} \quad (4.2.6)$$

Computer analysis [11] shows the system to be controllable and have eigenvalues of

$$\lambda_1 = 0.0$$

$$\lambda_2 = -.886$$

(4.2.7)

$$\lambda_3 = +.0045$$

$$\lambda_{4,5} = +.026 \pm j .644$$

indicating an open-loop-unstable system.

Before a design can begin, a suitable choice of output variables must be made. Intuitively one would control the rudder deflection with the yaw angle and the aileron deflection with the roll angle. Investigating the original regulator design made from minimizing a performance index [12] shows this primarily to be the case, where the respective angular velocities also show high gains in the corresponding feedback path.

In light of this information, the angles ψ and ϕ are chosen as the outputs of the system. These angular velocities could instead have

been chosen, but the all integrator block diagram of the plant (Figure 4.2.1) shows this would result in some unobservable states and would regulate the angular velocities and not the angular positions of the aircraft.

Computer analysis [11] shows the system to be observable and routinely yields a transfer function description of the plant as

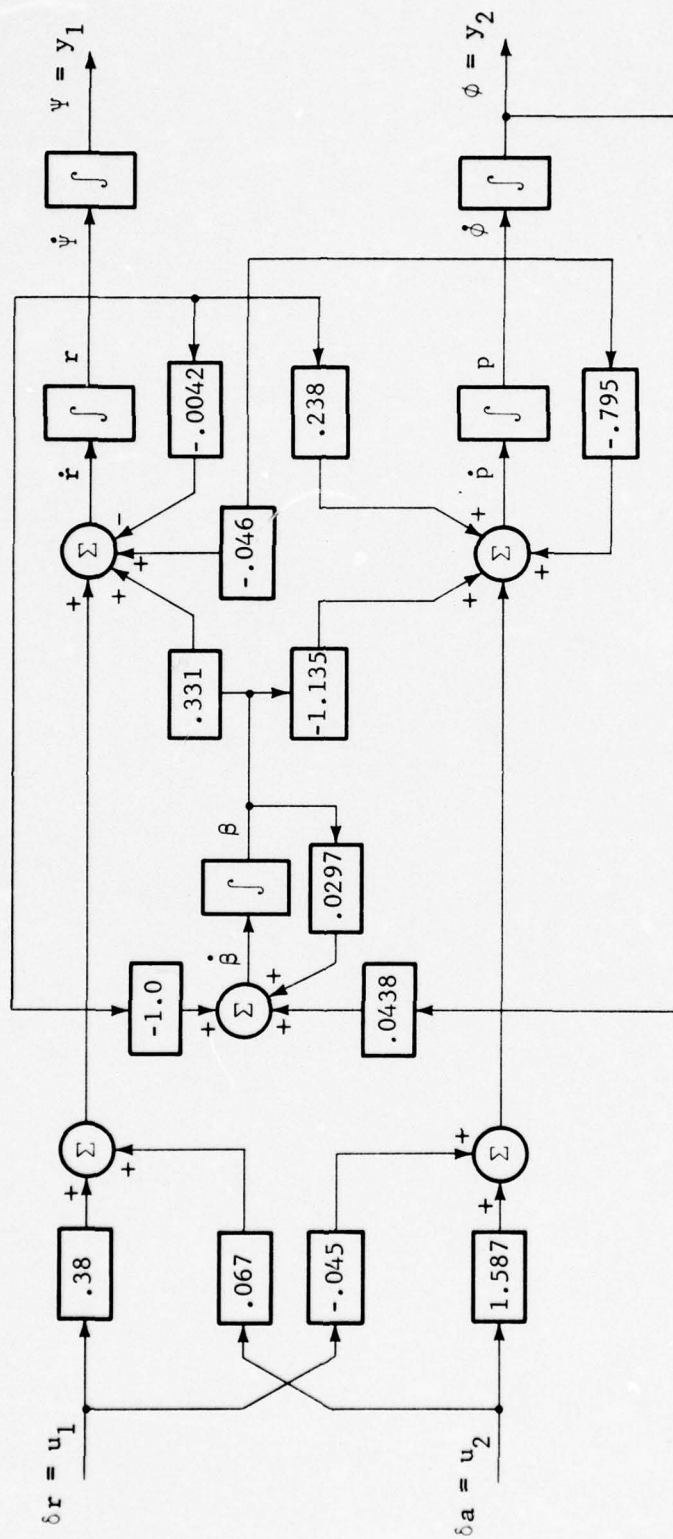
$$\begin{bmatrix} Y_1 \\ Y_2 \end{bmatrix} = \frac{1}{\text{DET}} \begin{bmatrix} -[.3807 s^3 + .3121 s^2 + .00893 s + .0195] & -s[.0404 s^2 + .0503 s + .447] \\ [.06715 s^3 - .0178 s^2 - .000588 s + .0226] & s[1.587 s^2 + .0624 s + .602] \end{bmatrix} \times \begin{bmatrix} \delta r \\ \delta a \end{bmatrix} \quad (4.2.8)$$

$$\text{where DET} = s[s^4 + .829 s^3 + .364 s^2 + .365 s - .00166] \quad (4.2.9)$$

4.3. Autopilot Main Program

Before the main program for the autopilot design may be run, two files must be created on disk. The first is file Q1.DAT, which is used to store the Q values from COMMON /A4/, if desired. The integer ϕ is stored in the second space of the first line, and the file is left with no line numbers. This procedure is illustrated in Appendix 1.

The second file which must be created is the data file DATA12.FOR. This contains the data subroutine DATA12, which initializes the system data



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Figure 4.2.1. All integrator block diagram showing state space description of autopilot plant with BLTPLTS.

in COMMON /A1/ and the programming variables in COMMON /A2/ via standard FORTRAN assignment statements. A copy of DATA12.FOR is included in Appendix 2, along with all the autopilot programming routines.

The main program is executed from the command file MAIN.CMD as illustrated in Appendix 1. Upon its execution, the graphics terminal and supportive software are initialized, IEND is calculated, and subroutine DATA12 is called for the necessary data.

The subroutine QDSKRD is then called. If disk file Q1.DAT has \hat{Q} values stored in it from the previous run, these values are automatically reloaded into arrays QR and QI in COMMON /A4/ and the routine returns the integer value IDUM = 1. This causes the message

Q VALUES WERE OBTAINED FROM DISK

to be typed on the terminal screen and execution proceeds to one of the six main program options.

If no Q values are stored from the previous run, QDSKRD returns the integer value IDUM = \emptyset . This causes matrix XS to be set equal to the identity matrix, subroutine MAP (\emptyset) is called, causing the value of $\overline{G(s)}$ along the jw axis to be stored in COMMON /A32/. Subroutine MIMP is called, taking these values and placing them in COMMON /A4/ also.

The subroutine QINV is then called. This subroutine takes the inverse of the (2x2) square complex matrices, IEND in number, stored in COMMON /A4/. It returns the inversed values again to COMMON /A4/. Note that one could have started with the inversed system data in routine DATA12.

Taking the point by point inverse with QINV saves one the labor of computing the inverse transfer function matrix by hand. Subroutine QINV is stored on disk file QINV2.FOR and is included in Appendix 2.

In either case, COMMON /A4/ now stores the open loop system frequency data. The message

TYPE: 0 TO STOP

TYPE: 1 FOR OPEN LOOP PACKAGE

TYPE: 2 FOR INPUT COMPENSATOR PACKAGE

TYPE: 3 FOR CLOSED LOOP PACKAGE

TYPE: 4 FOR OUTPUT

TYPE: 5 TO MODIFY INTEGER COMMON /A2/ VALUES

is typed on the screen. Entering the desired integer causes execution to proceed to a particular conversational routine. Returning from one of these routines causes this same message to appear and allows the user to chose a different option, until he decides to stop.

4.4. The Interactive Design Packages

Entering ϕ to stop causes subroutine FINISH to be called. Similarly entering 4 calls routine OUT and entering 5 calls routine MODIA2. These subroutines are included with the other five auxiliary type subroutines in the file AUX FOR.

In this same manner the six computationed routines have been copies from their respective .SAV file into the file COMP.FOR. The seven

graphics routines are similarly stored in GRAPH FOR. The necessary common block usage, dimensioned for the autopilot example, is included in each subroutine of these three files.

4.4.1. Subroutine OLISP

Typing 1 for the open loop package calls the subroutine OLISP. This routine is useful in investigating the diagonal dominance of the plant and in the design of constant input compensators. It can also be used to show the effect of actuator failure at an input to the plant.

The routine has a functional flow as diagrammed in Figure 4.4.1. Upon entering the routine, work matrix XS is set equal to input compensating matrix XKA. The open loop system Q variables are stored in arrays XR1 and XII in COMMON /A32/ and subroutine GERSH computes the Gershgorin bands for those values. The message

DEALING W/ OPEN LOOP INPUT CONST. COMP

is typed on the screen to remind the user of the action being taken. The routing message

TYPE \emptyset TO RETURN, 1 TO PLOT, 2 TO MODIFY

then appears on the terminal screen. Entering 2 calls subroutine MODXS for modification and/or intermediate output of the constant input compensator and upon its return the same directioning message is typed on the screen. Entering \emptyset to return sets system matrix XKA equal to the routine's work matrix XS, and causes the execution to continue at the main program.

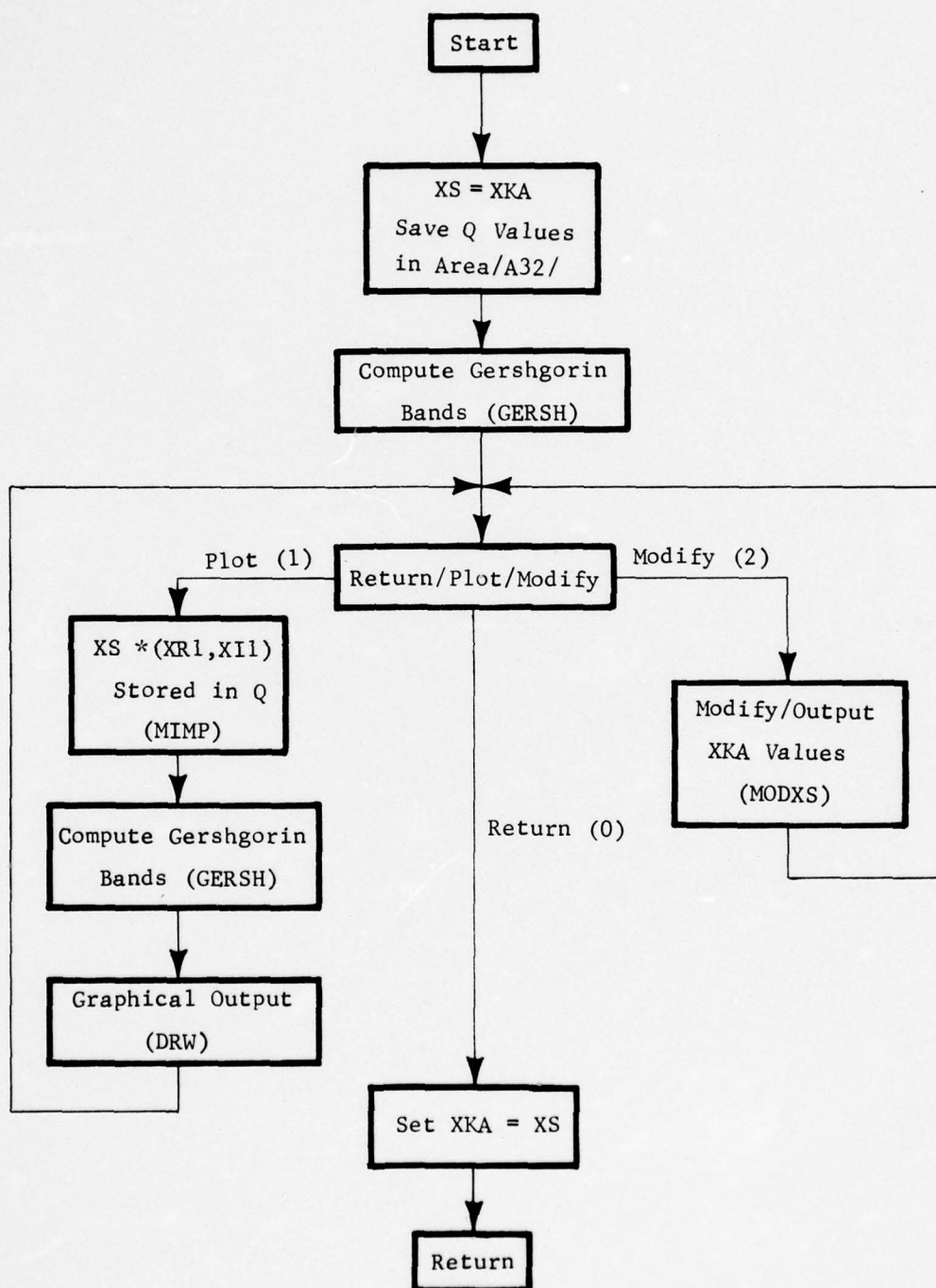


Figure 4.4.1. Functional flow chart of subroutine OLISP.

If instead one chooses to display a graphical output by entering 1, the following happens. Subroutine MIMP takes the matrix product of matrix XS and the original \hat{Q} , values stored in COMMON /A32/ and places the result in arrays QR and QI in COMMON /A4/. Subroutine GERSH computes the moduli of the new \hat{Q} values and subroutine DRW is called at entry point DRW1 for a graphical display as described in Section 3.3.2. After all the desired plots are inspected, the routing message

TYPE: 0 TO RETURN, 1 TO PLOT, 2 TO MODIFY

again appears on the screen, allowing further modification and plotting, or a return to the main program.

When XS is set equal to the identity matrix, the information displayed is the open loop plant. Otherwise it is the compensated plant.

The important thing to note is that when returning to the main program, the values of \hat{Q} are those of the last input compensator used. If it is not desired to change these values from what they were when the package was entered, matrix XS should be set equal to the identity matrix and the results plotted once. The effect of successive changes in XS are not accumulative unless a return is made to the main program.

One may exploit this particular quirk as follows. Let the plant \hat{Q} values be stored on disk. This is usually the case at this stage in a design. Enter this design package and set matrix XS equal to the identity matrix, except for one diagonal entry. This is set to zero to represent an actuator failure at that input. Return to the main program, then return to this open loop design package and set XS equal to the desired design value if the input compensator. The resulting frequency loci will be those when this actuator has failed. When compensator XKA is not diagonal, this

type of failure may have astounding effects on the resulting stability region of the closed loop system. Note that the \hat{Q} values are partially destroyed by this type of procedure, and therefore should be saved beforehand.

4.4.2. Subroutine IFDCD

Typing 2 for the input compensator package option in the main program calls subroutine IFDCD. This routine is very similar to the open loop package OLISP. The difference between the two is that here the input compensator is allowed to be a function of frequency. It therefore finds its usefulness in those types of designs, but it may be used for investigation of the plant frequency loci or show the effect of actuator failures as in subroutine OLISP. A functional flow chart of routine IFDCD is included as Figure 4.4.2.

Upon entering this package, the systems \hat{Q} values from COMMON /A4/ are stored in arrays XR2 and XI2 in labeled common area /A33/. The work matrix XC is set equal to the input frequency dependent matrix XKB and subroutine GERSH is called to compute the sum of the moduli of the Q values. The message

DEALING W/ FREQ. DEPNDT. INPUT COMP.

is typed on the screen.

The user then has the same three options as in the previous routine as the routing message

TYPE: \emptyset TO RETURN, 1 YO PLOT, 2 TO MODIFY

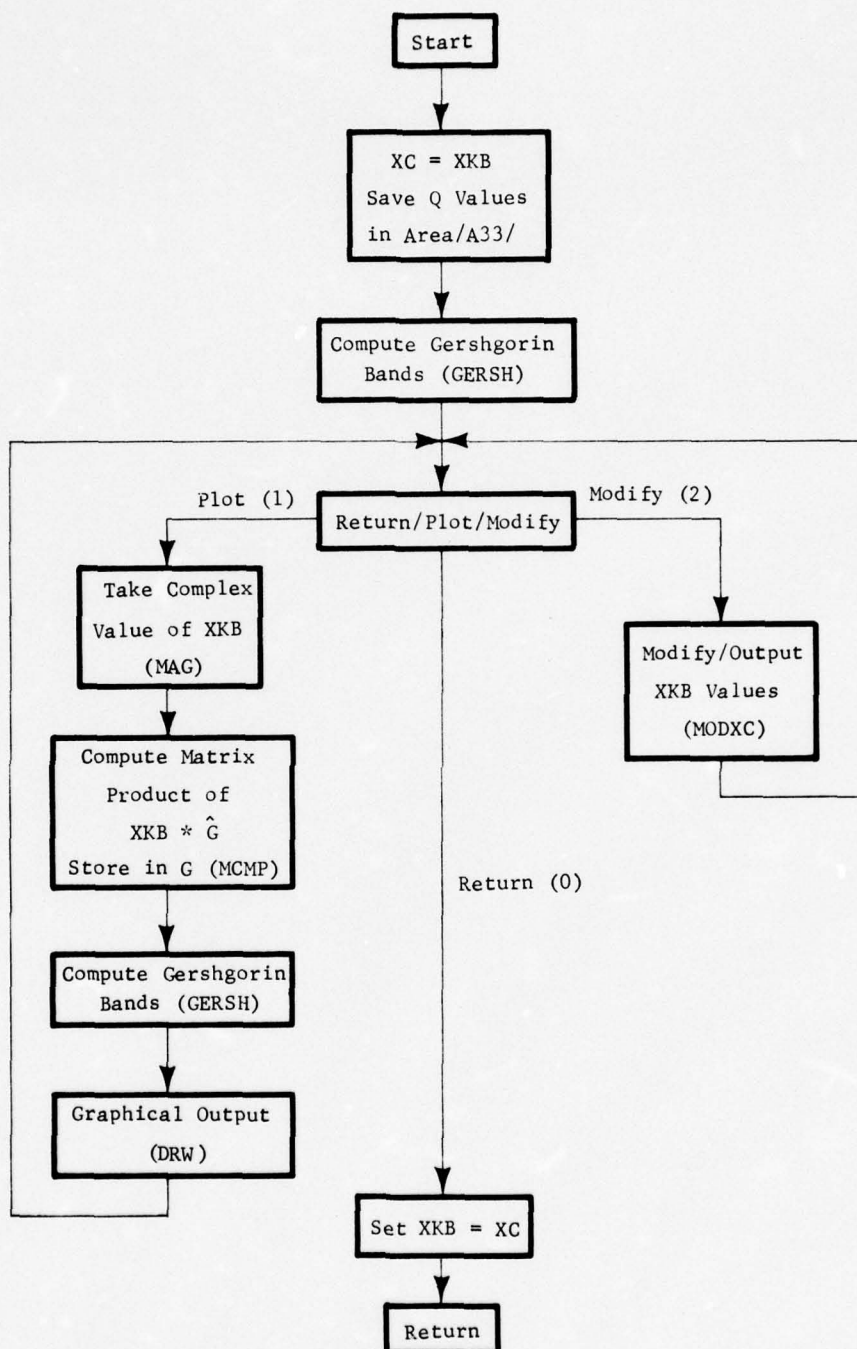


Figure 4.4.2. Functional flow chart of subroutine TFDGP.

appears on the terminal screen. Entering 2 calls subroutine MODXC to modify values in the inverse transfer function of the input compensator and/or cause its intermediate output to be performed. Upon completion, the same routing message is typed on the screen. Entering \emptyset to return sets array XKB equal to the work array XC and causes a return to the main program.

Entering 1 to plot causes the following to happen. Subroutine MAG takes the complex value of the input compensator XKB stored in array XC and subroutine MCMP performs the product of these complex values with those values in \hat{Q} at the entrance of the routine, stored in COMMON /A33/. This matrix product is stored in the \hat{Q} arrays in COMMON /A4/. Subroutine GERSH then computes the moduli of the compensated plant and subroutine DRW is called at entry point DRW1 for a graphical display of the computed information. After all the desired plots are inspected, the original routing message appears on the screen allowing for further modification or a return to the main program.

The effects of repetitive changes in array XC are not accumulative and those values of the system stored in the \hat{Q} arrays in COMMON /A4/ are for the last input compensator used. Caution should again be exercised when leaving the package and entering another package from the main program.

4.4.3. Subroutine CLSP

Typing 3 for the closed loop package option in the main program calls subroutine CLSP. This programming segment allows for the interactive modification of the feedback matrix, \bar{F} , the computation of the Ostrowski bands, and the graphical display of the resulting closed loop composite

inverse Nyquist diagrams using the maximum possible screen size. The functional flow chart of this package is included as Figure 4.4.3.

When entering the routine, work matrix XS is set equal to the value of the feedback matrix F from COMMON /A1/. Subroutine GERSH is called to compute the Gershgorin bands and subroutine OSTROW computes the value of the Ostrowski band radii from the Gershgorin bands. The message

DEALING W/ CLOSED LOOP SYSTEM

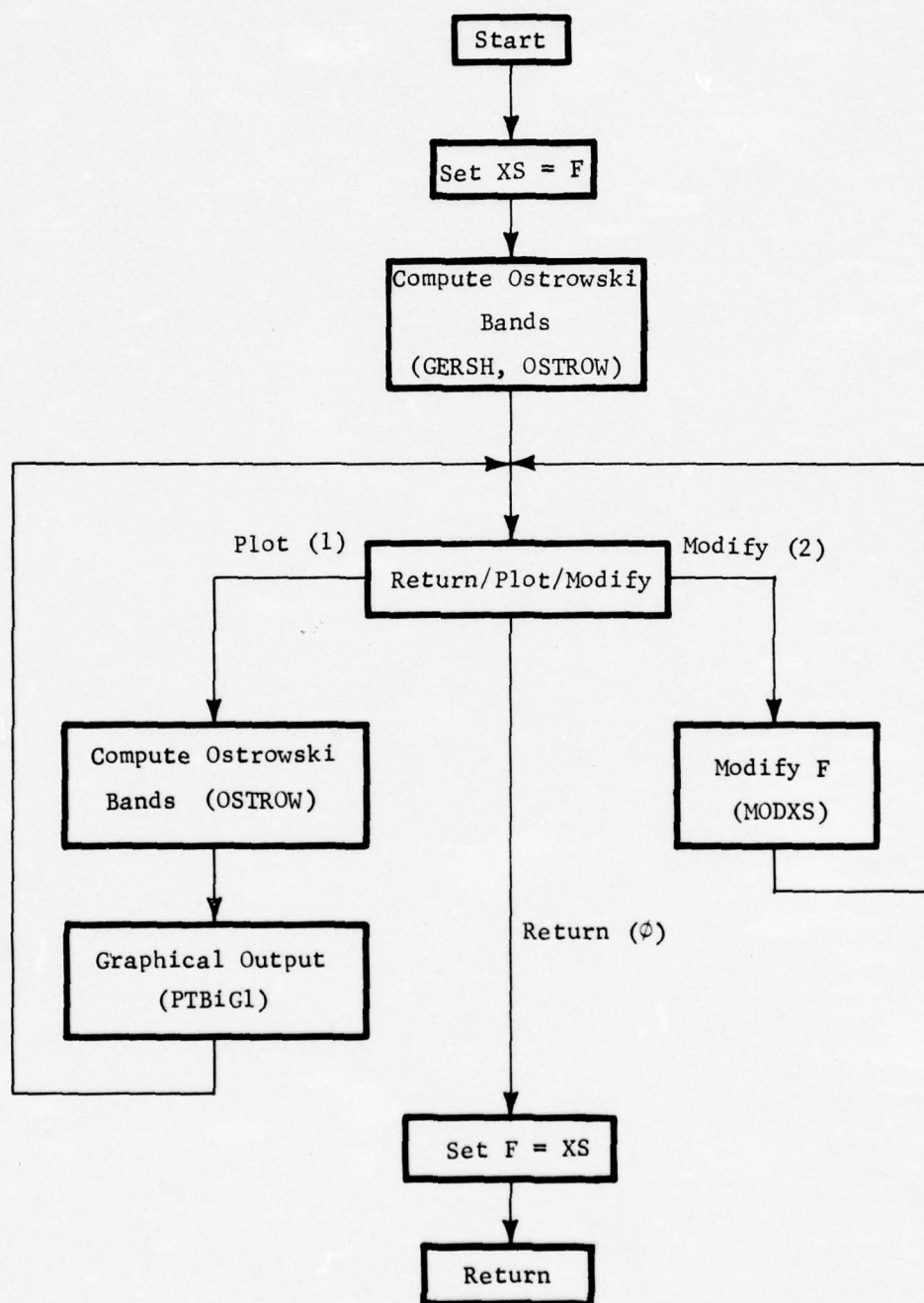
is typed on the terminal screen to remind the user that the matrix XS has taken on the function of the feedback matrix \bar{F} .

The routing message

TYPE: \emptyset TO RETURN, 1 TO PLOT, 2 TO MODIFY

appears on the screen. Entering \emptyset sets matrix F equal to work matrix XS and a return is made to the main program. Entering 2 to modify, results in the calling of subroutine MODXS for changing values in the feedback matrix and/or their output. The return from routine MODXS causes the original routing message to again appear on the terminal screen.

If a graphical output is desired the integer 1 is entered. This causes subroutine OSTROW to compute the Ostrowski bands and subroutine PTBIG1 is called for the graphical investigation of the diagonal closed loop inverse frequency loci. Returning from the graphics routine causes the original routing message to be typed on the screen for further modification of feedback gains or a return to the main program for a different design option to be executed.



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Figure 4.4.3. Functional flow chart for CLSP.

This package is also useful in showing the effects of transducer failures. By changing a specific feedback element from its design value to zero with the interactive modifying routine MODXS, the composite frequency loci of the system with this failure may be investigated.

4.5. Discussion of Autopilot Example

The autopilot plant transfer function (4.2.8) and (4.2.9) and the programming variables in COMMON /A2/ were assigned in data subroutine DATA12. Constant input compensator XKA was initialized to the identity matrix as well. The open loop package OLISP was used to investigate the composite frequency loci of the plant.

The plant without constant matrix compensation is observed to be highly nondominant (Figures 4.5.1-4.5.7). Several different constant input compensator matrices were constructed in order to achieve diagonal dominance of the open loop system. None of these attempts were successful in achieving a diagonally dominant open loop system at all frequencies.

In fact, following Hankins [13] it can be shown that the autopilot plant cannot be put into a diagonally dominant form with a constant matrix compensator.

One attempts to obtain row diagonal dominance, as this has certain advantages when matrix compensation is placed in the forward path [14] and the system was originally constructed so this would be the case. Therefore \hat{K} is partitioned by rows and each of these rows is designed by considering the row dominance in that corresponding row of \hat{Q} . This requires $\hat{G}(j\omega)$ to

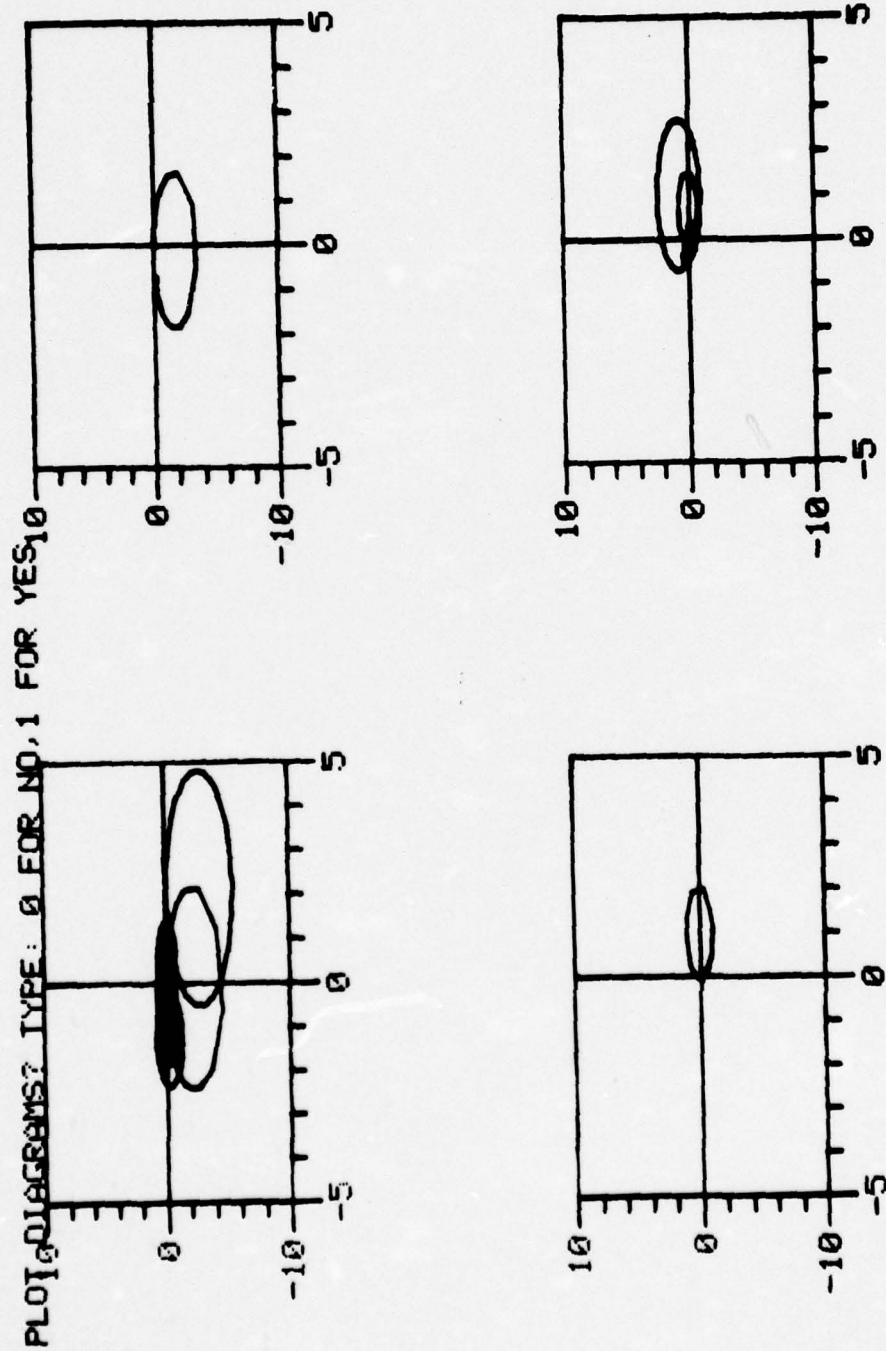


Figure 4.5.1. The four inverse Nyquist diagrams of the plant for $0 < \omega < .8$.

1 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR ROWS, 2 FOR COLUMNS

PLOT₂ DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

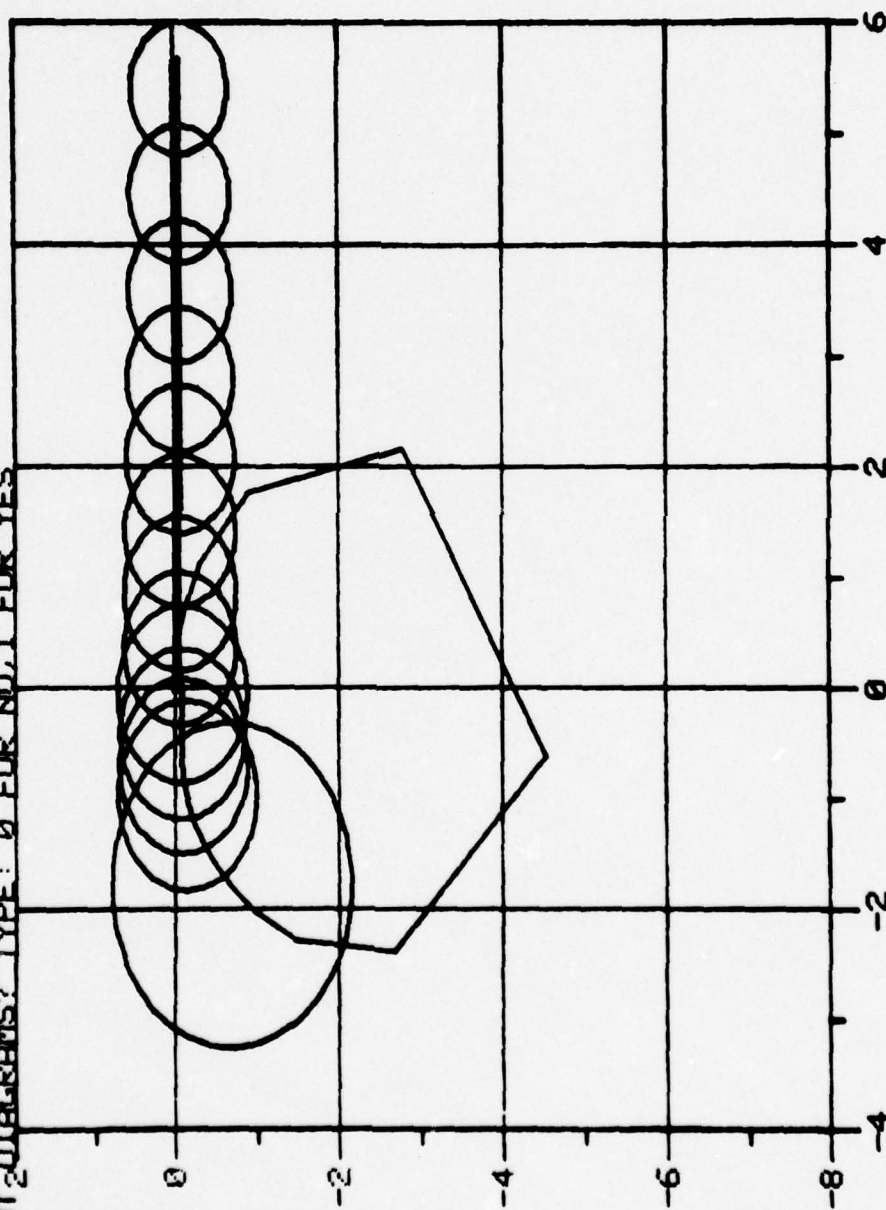


Figure 4.5.2. \hat{g}_{11} with Gershgorin bands for $0 < \omega < 1.6$.

0 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR ROWS, 2 FOR COLUMNS

PLOT DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

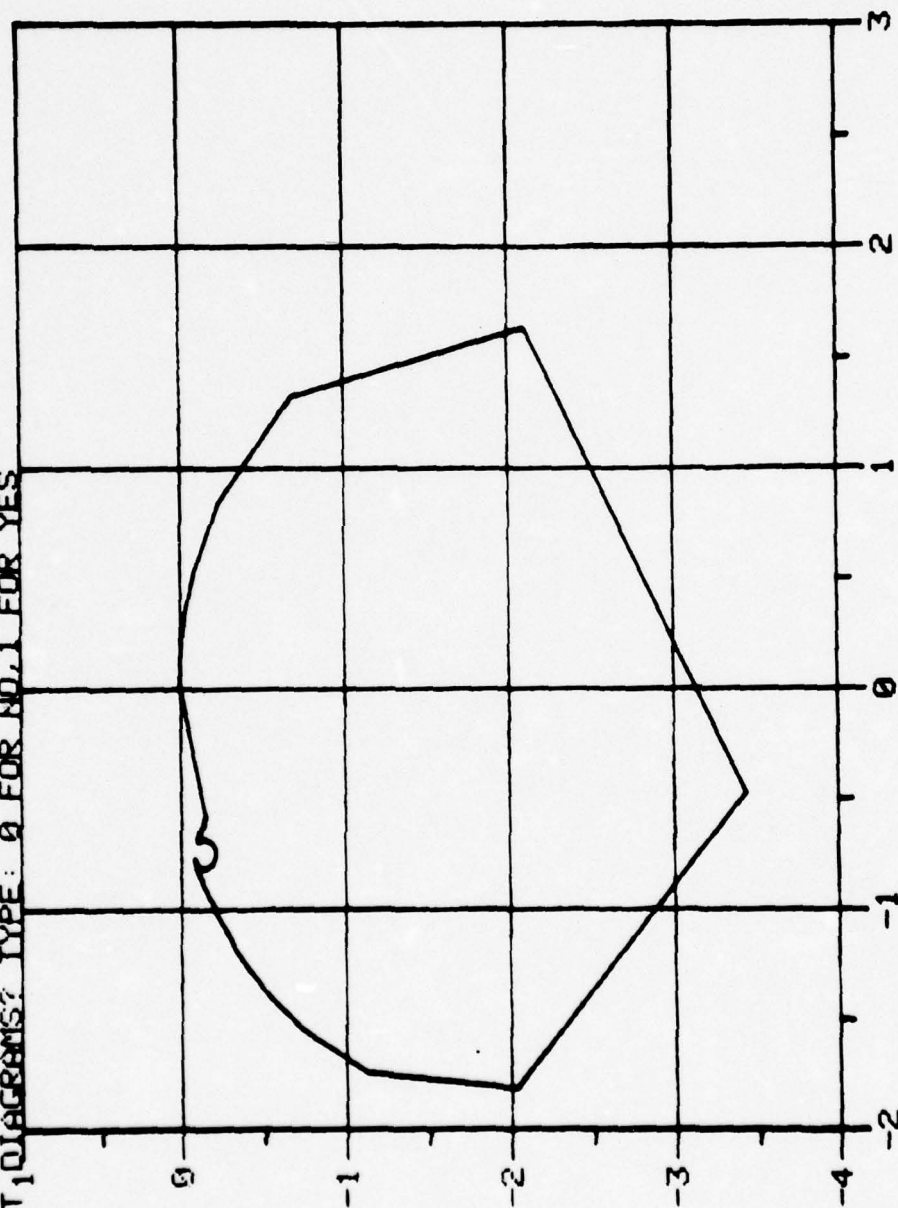


Figure 4.5.3. \hat{g}_{12} for $0 < \omega < 1.6$.

0 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR ROWS, 2 FOR COLUMNS

PLOT DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

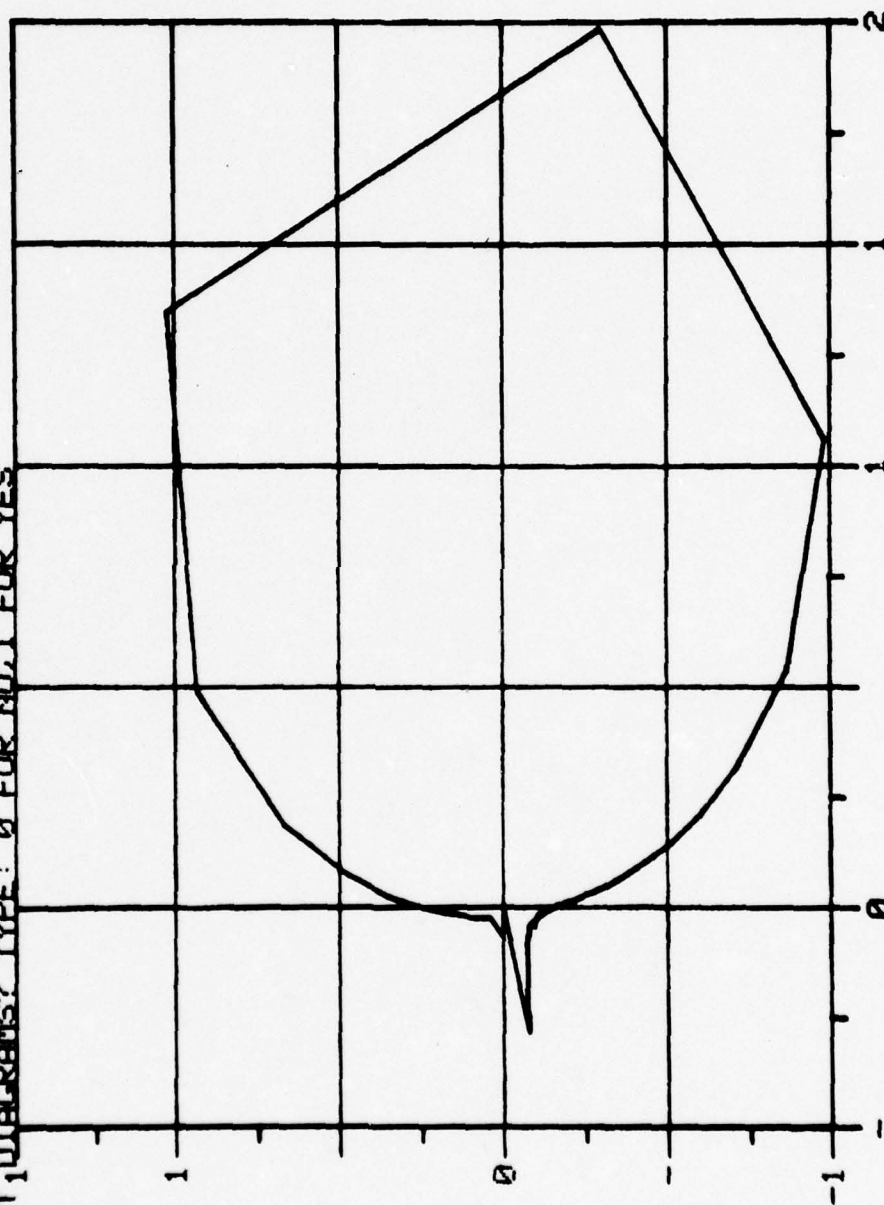


Figure 4.5.4. \hat{g}_{21} for $0 < \omega < 1.6$.

1 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR ROWS, 2 FOR COLUMNS
 PLOT DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

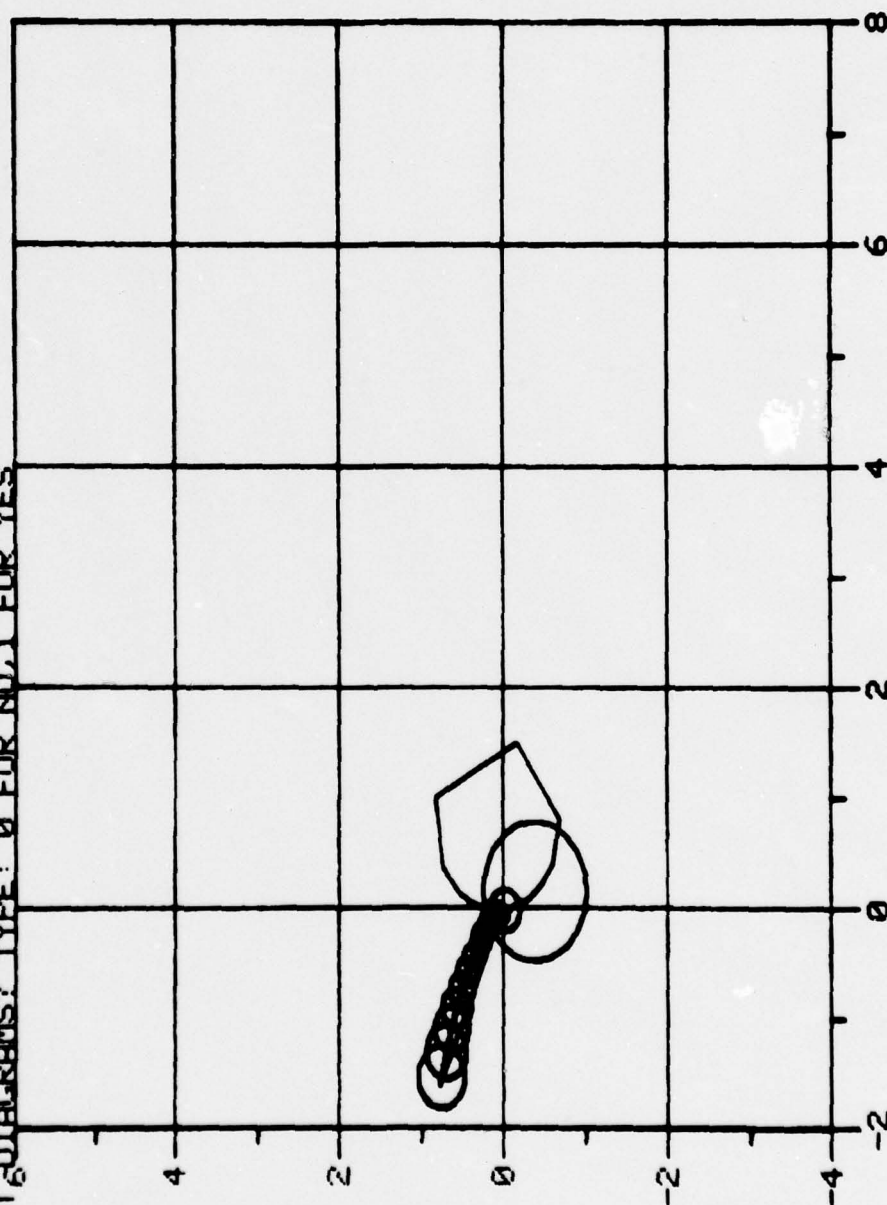


Figure 4.5.5. \hat{g}_{22} with Gershgorin bands for $0 < \omega < 1.6$.

1 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR ROWS, 2 FOR COLUMNS

PLOT DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

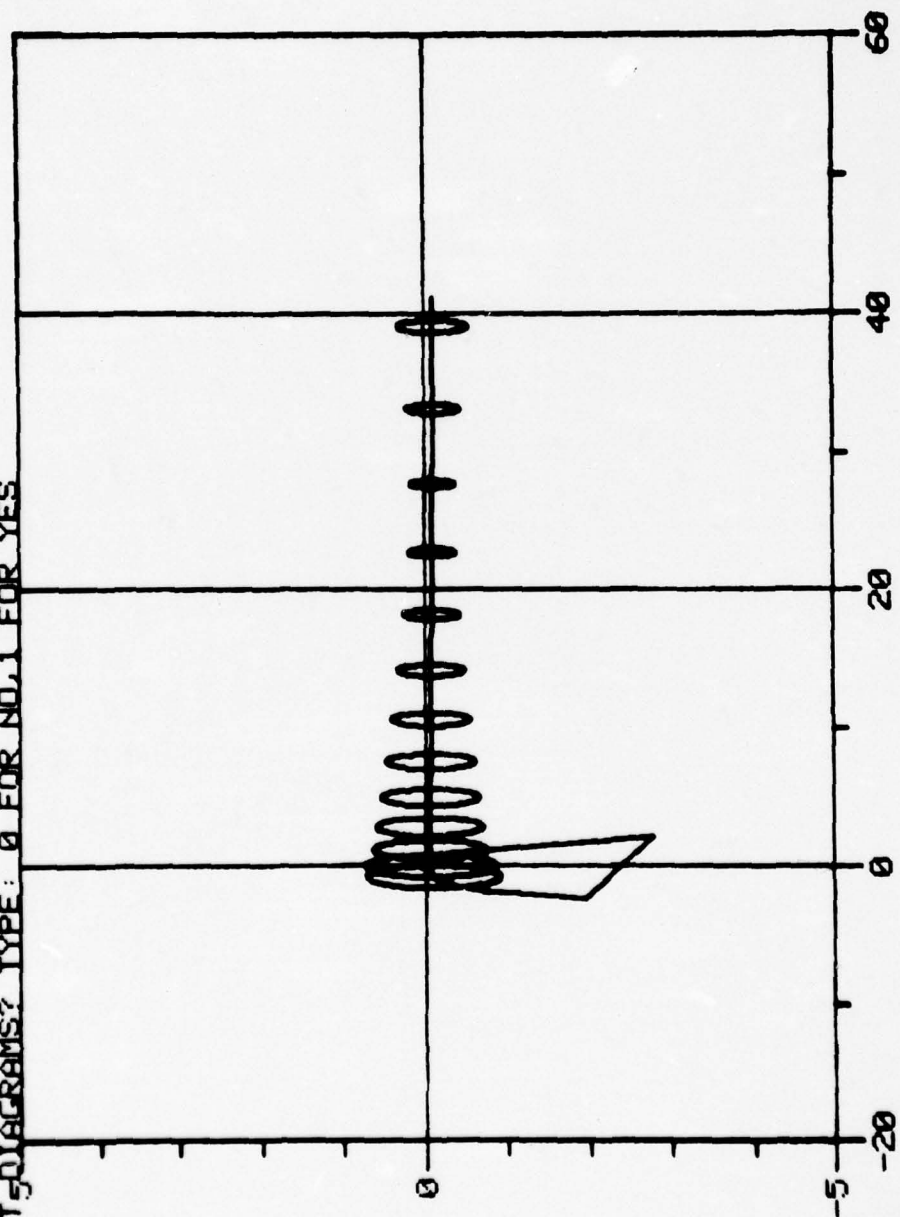


Figure 4.5.6. \hat{g}_{11} with Gershgorin bands for $0 < \omega < 4.0$.

1 PLOT BANDS? TYPE: 0 FOR NO, 1 FOR YES, 2 FOR COLUMNS

PLOT DIAGRAMS? TYPE: 0 FOR NO, 1 FOR YES

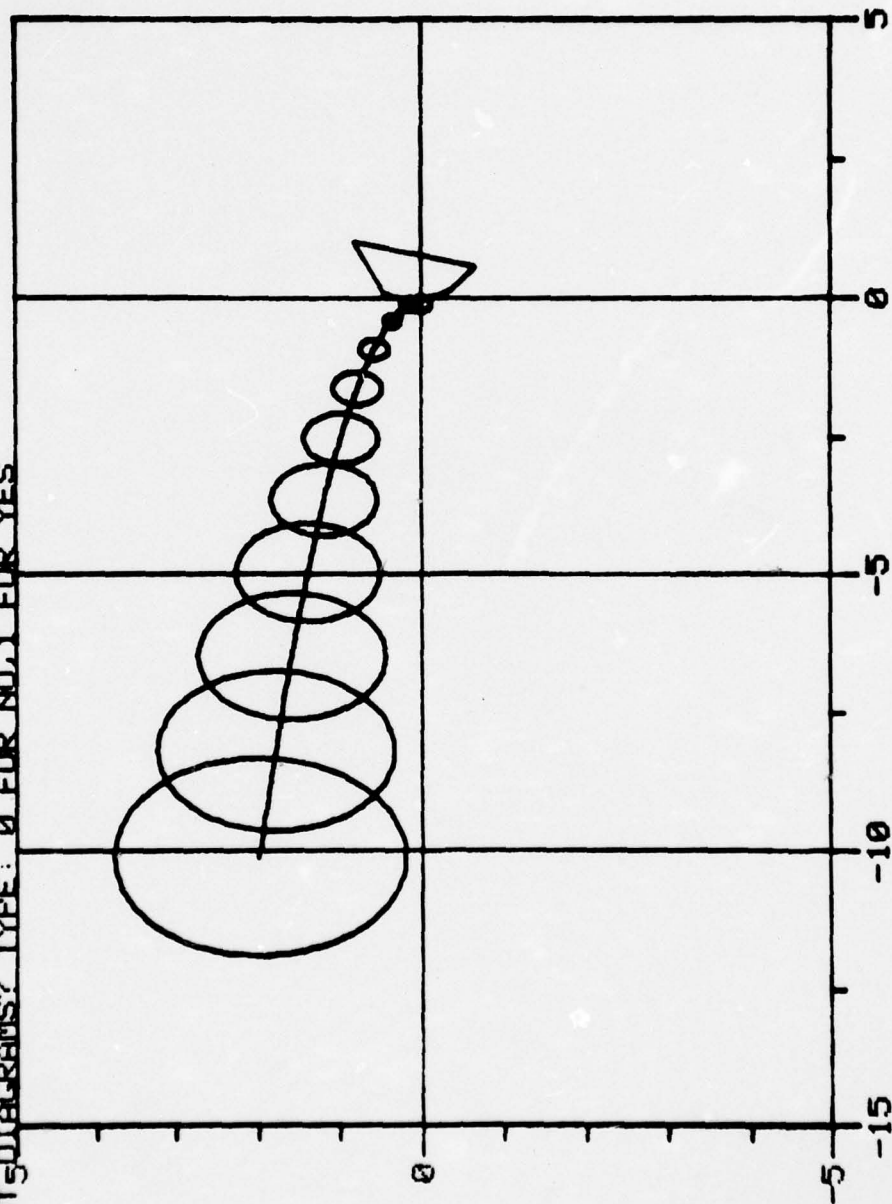


Figure 4.5.7. \hat{g}_{22} with Gershgorin bands for $0 < \omega < 4.0$.

be partitioned by columns as

$$\hat{G}(j\omega) = [\alpha_1 + j\beta_1 : \alpha_2 + j\beta_2]. \quad (4.5.1)$$

Hawkins shows that diagonal dominance cannot be obtained if the two eigenvalues of the matrix

$$\bar{C}_1 = [\alpha_1 \alpha_1^T + \beta_1 \beta_1^T - \alpha_2 \alpha_2^T - \beta_2 \beta_2^T] \quad (4.5.2)$$

have the same sign. He then carries this idea further to a generalized graphical method of obtained dominance for a two input two output plant by specifying numerical bounds for the elements in such a constant compensating matrix.

For the autopilot plant, the value of the inverse transfer function at $j\omega = .1$ is

$$\hat{G}(j.1) = \begin{bmatrix} -(2.40 + j1.95) & -(1.48 + j1.73) \\ (.74 - j.94) & (.56 - j.66) \end{bmatrix}. \quad (4.4.3)$$

Although \hat{q}_{11} is already both row dominant and column dominant for this frequency, C_1 is computed from (4.5.2) as

$$\bar{C}_1 = \begin{bmatrix} 3.18 & -1.05 \\ -1.05 & .608 \end{bmatrix}. \quad (4.4.4)$$

Observe that \bar{C}_1 is symmetric by construction and that $\bar{C}_2 = -\bar{C}_1$. The eigenvalues of \bar{C}_1 are found to be +3.47 and +.32. As they have the same sign, diagonal dominance in this manner can never be achieved by a constant compensator \hat{K} .

As open loop diagonal dominance cannot be obtained with constant compensation, a design for the autopilot regulator problem could proceed in one of the following directions:

1) Nothing prevents one from attempting to construct a nondiagonal input matrix compensator, which is a function of frequency, to obtain an open loop diagonally dominant system. However, one should keep in mind the necessary complexity of the control scheme which will result after the additional diagonal compensators for the desired input-output time domain responses are included. Present controlling methods in actual use should be surveyed to see if these means are justifiable in terms of maintenance and reliability. The resulting improvement in performance of the operating system as a function of control will ultimately be the deciding factor. Clearly more fundamental work in achieving open loop dominance via these means is needed.

2) Off diagonal elements which are a function of frequency may be inserted in the feedback path to allow for transient compensation of the interaction of the transfer functions in the closed loop system. As the inverse Nyquist method has been used only in industrial control problems which have not needed this device, all the consequences of this type of control are not fully known.

3) We chose as outputs from the state description of the plant the yaw angle and the roll angle. These motions are highly coupled through the dynamics of the aircraft with the sideslip angle. An analysis should be performed to see how the sideslip angle could be included as an output so that the resulting description of the inverse plant has a dominant form.

Unfortunately, an algorithm for such an analysis has not yet been determined.

4) It should be kept in mind that one was given highly accurate but typical values in the perturbed stated equations about an operating point for the plant description. These resulted in frequency loci which were somewhat atypical in that they exhibited peculiar behavior in certain narrow frequency regions. The validity of this precise energy description of the plant should be questioned. It of course, would be better to measure the frequency description of the actual plant in operation. This could be done at various altitudes and velocities, and under different loading conditions. Perhaps a more meaningful place to begin a design would be an investigation of the change in the input-output frequency description under such different operating conditions.

5. CONCLUDING REMARKS

The inverse Nyquist array method represents an extremely practical way in which one may solve multivariable control design problems. It is somewhat limited in that the number of system inputs and outputs must be equal. A large number of plants fit into this category, however.

Certainly an important aspect of the method is the great freedom the designer has in formulating his problem. This is not only from a standpoint of the structure of the controlling scheme. System specifications in terms of marginally acceptable performance may be included in the event of controller variations on total failure or transducer and actuator hardware. Plant variations may be dealt with in a straightforward manner.

The results presented here deal only with a linear and time invariant plant. Some theory for the nonlinear and time varying case based on describing functions does exist [2, pp. 179-184]. These results could probably be extended to sample data systems as well.

The somewhat intuitive action of the design method naturally calls for an interactive computer with graphics facility for its solution. The recent development and present availability of this hardware makes such an alternative method possible. In addition, it sheds a new light on graphically oriented design procedures in general.

This method has been applied to a variety of practical applications [2, pp. 190-219]. As it becomes more widely accepted, results will come forth allowing for additional theoretical work. In fact, very recent developments allow the diagonal dominance criterion to be considerably loosened; the need

for detailed high frequency performance in the inverse plant has been eliminated as well [14]. Without doubt, further work using this alternative type of approach will yield practical solutions to many complex design problems in the near future.

REFERENCES

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APPENDIX 1

The data file Q1.DAT has a zero in the second space of the first line and may be made using the CREATE command as follows:

```
CREATE Q1.DAT
Create: DSKC:Q1.DAT[404,443]
00100    0$
*ES
[DSKC:Q1.DAT[404,443]]
```

```
.
TYPE Q1.DAT
0
.
```

The main program stored in disk area [404,443] may be executed from the monitor command

EXE@MAIN.CMD

where the command file MAIN.CMD is

```
AUTOMP.FOR
,OLISP.FOR
,IFDCP.FOR
,CLSP.FOR
,COMP.FOR
,GRAPH.FOR
,AUX.FOR
,DATA12.FOR
,QINV2.FOR
,SYS:AG210/SEA
```

This procedure creates relative binary files of 'FILENAME'.REL as
the following:

```
EXE@MAIN.CMD  
  
FORTRAN: AUTOMP  
MAIN.  
FORTRAN: OLISP  
OLISP  
FORTRAN: IFDCP  
IFDCP  
FORTRAN: CLSP  
CLSP  
FORTRAN: COMP  
MAP  
MAG  
MIMP  
MCMP  
GERSH  
OSTROW  
MMLJ  
FORTRAN: GRAPH  
PTNY1  
PTGR1  
PTOST1  
PTBIG1  
MAK4  
DRW  
FORTRAN: AUX  
MODIA2  
MODXS  
MODXC  
QOUT  
QDSKWR  
QDSKRD  
OUT  
FINISH  
FORTRAN: DATA12  
DATA12  
FORTRAN: QINV2  
QINV
```


If these relative binary files are already created and data is stored in disk file Q1.DAT, the execution of the main program and its termination appears as

```
EXE@MAIN.CMD
LINK: Loading
[LNKXCT AUTOMP Execution]
Q VALUES WERE OBTAINED FROM DISK
TYPE: 0 TO STOP
TYPE: 1 FOR OPEN LOOP PACKAGE
TYPE: 2 FOR INPUT COMPENSATOR PACKAGE
TYPE: 3 FOR CLOSED LOOP PACKAGE
TYPE: 4 FOR OUTPUT
TYPE: 5 TO MODIFY INTIGER COMMON/A2/VALUES
```

0

```
YOU ARE DONE: BEST PRINT OUT SYSTEM!!!
FOR OUTPUT TYPE:
  0 FOR NONE, 1 FOR TTY, 2 FOR LP
```

0

```
STORE Q VALUES? TYPE 0 FOR NO, 1 FOR YES
```

0

```
5: @STOP
```

```
END OF EXECUTION
CPU TIME: 3.80 ELAPSED TIME: 1:30.37
EXIT
```

APPENDIX 2

The twenty-one subroutines are stored individually on disk in area [404,443] in file names corresponding to the subroutine name with extension of SAV. They have been left unprotected and therefore may be copied into a user's disk area by executing the monitor command

COPYXXXXXX.SAV[404,443]

where XXXXXX is the appropriate filename. One may copy all twenty one routines at once by using the option

COPY*.SAV[404,443]

In a similar fashion the programming for the autopilot problem may be transferred to a different users area. These files include:

I. COMP.FOR which contains the computational routines:

- 1) Subroutine MAP, which computes points on the frequency loci for any region on the Nyquist contour D.
- 2) Subroutine MAG, which computes the complex value of any transfer function matrix.
- 3) Subroutine MIMP, which takes the product of a real matrix and a complex matrix stored by real and imaginary parts.
- 4) Subroutine MCMP, which takes the product of two complex matrices each stored by real and imaginary parts.
- 5) Subroutine GERSH, which computes various sums of the moduli of elements in the open loop plant.

- 6) Subroutine OSTROW, which computes the radi of the Ostrowski bands in the closed loop system.
- 7) Subroutine MMIJ, which searches for minimum and maximum values on a composite frequency locus.

See Table A.2.1 for a listing.

II. GRAPH.FOR which contains the graphics routines:

- 1) Subroutine PTNY1, which plots a single frequency loci on the screen.
- 2) Subroutine PTGR1, which plots the Gershgorin bands based on rows or columns on the screen.
- 3) Subroutine PTOST1, which plots the Ostrowski bands based on rows or columns on the screen.
- 4) Subroutine PTBIG1, which interactively causes a single composite Nyquist diagram with Ostrowski bands using the maximum possible screen size to be plotted.
- 5) Subroutine MAK4, which plots four separate frequency loci with Gershgorin bands on the screen simultaneously.

A listing of this program is given in Table A.2.2.

III. AUX.FOR which contains the auxiliary routines:

- 1) Subroutine MODIA2, which interactively modifies the range of the frequency loci and the number of Gershgorin or Ostrowski bands.
- 2) Subroutine MODXS, which interactively modifies and/or outputs scalar matrices.
- 3) Subroutine MODXC interactively modifies and/or outputs frequency dependent compensators.


```

00100 C *****
00200 C *****
00300 SUBROUTINE MAP(ICIRCL)
00400 C *****
00500 C TAKES COMPLY MAP OF G INTO (XR1,XI1)
00600 C ICIRCL = 0 FOR S BETWEEN OMEG0 AND OMEGND ONLY
00700 C          = 1 FOR S ON THE SEMICIRCULAR ARC ONLY
00800 C          = 2 OR ANY OTHER VALUE FOR ENTIRE NYQUIST PATH
00900 C *****
01000 IMPLICIT COMPLEX(C)
01100 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
01200 C *****
01300 C USES COMMON/A1/,COMMON/A32/ ALSO
01400 C *****
01500 COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XLB(2,2,3,2),
01600 1 XKA(2,2),XLA(2,2),F(2,2)
01700 COMMON/A32/XR1(2,2,201),XI1(2,2,201)
01800 C *****
01900 970 FORMAT(X,'A POLE AT OMEGA = ',F7.2,2X,'FOR G(',I1,',',
02000 11,',',I3,',')')
02100 NTIMES=(OMEGND-OMEG0)/DTOMEG)+1
02200 DO 780 I=1,MO
02300 DO 780 J=1,MO
02400 IF(ICIRCL.EQ.1) GO TO 751
02500 C MAPS TOP STRIP OF JW AXIS
02600 DO 750 N=1,NTIMES
02700 IF(N.GT.IEND) GO TO 780
02800 OMEGA=OMEG0+(N-1)*DTOMEG
02900 SIGMA=0.0
03000 GO TO 730
03100 720 SIGMA=-DTOMEG
03200 TYPE 970,OMEGA,I,J,N
03300 730 CSS=CMPLX(SIGMA,OMEGA)
03400 CNUM=(0.0,0.0)
03500 CDOM=(0.0,0.0)
03600 DO 740 K=2,6
03700 CNUM=CNUM+CMPLX(G(I,J,K,1),0.0)*CSS**(K-1)
03800 CDOM=CDOM+CMPLX(G(I,J,K,2),0.0)*CSS**(K-1)
03900 740 CONTINUE
04000 CNUM=CNUM+CMPLX(G(I,J,1,1),0.0)
04100 CDOM=CDOM+CMPLX(G(I,J,1,2),0.0)
04200 IF(CDOM.EQ.(0.0,0.0)) GO TO 720
04300 C=CNUM/CDOM
04400 XR1(I,J,N)=REAL(C)
04500 XI1(I,J,N)=AIMAG(C)
04600 750 CONTINUE
04700 IF(ICIRCL.EQ.0) GO TO 780
04800 751 IF(ICIRCL.EQ.1) NTIMES=0
04900 C MAPS SEMICIRCULAR RADIUS PART
05000 DO 760 N=1,99
05100 N1=N+NTIMES
05200 OMEGA=OMEGND*COS(N*.0314159)
05300 SIGMA=OMEGND*SIN(N*.0314159)
05400 CSS=CMPLX(SIGMA,OMEGA)
05500 CNUM=(0.0,0.0)
05600 CDOM=(0.0,0.0)
05700 DO 742 K=2,6
05800 CNUM=CNUM+CMPLX(G(I,J,K,1),0.0)*CSS**(K-1)
05900 CDOM=CDOM+CMPLX(G(I,J,K,2),0.0)*CSS**(K-1)
06000 742 CONTINUE

```



```

06100      CNUM=CNUM+CMPLX(G(I,J,1,1),0.0)
06200      CDOM=CDOM+CMPLX(G(I,J,1,2),0.0)
06300      IF(CDOM.EQ.(0.0,0.0))TYPE 970,OMEGA,I,J,N
06400      IF(CDOM.EQ.(0.0,0.0))GO TO 760
06500      C=CNUM/CDOM
06600      XR1(I,J,N+NTIMES)=REAL(C)
06700      XI1(I,J,N+NTIMES)=AIMAG(C)
06800      CONTINUE
760      IF(ICIRCL.EQ.1)GO TO 780
C      MAPS BOTTOM STRIP OF JW AXIS
07100      DO 770 N=1,NTIMES
07200      N2=N+99+NTIMES
07300      OMEGA=OMEGND-(N-1)*DTOMEG
07400      SIGMA=0.0
07500      GO TO 733
07600      723      SIGMA=-DTOMEG
07700      TYPE 970,OMEGA,I,J,N2
07800      733      CSS=CMPLX(SIGMA,OMEGA)
07900      CNUM=(0.0,0.0)
08000      CDOM=(0.0,0.0)
08100      DO 743 K=2,6
08200      CNUM=CNUM+CMPLX(G(I,J,K,1),0.0)*CSS**(K-1)
08300      CDOM=CDOM+CMPLX(G(I,J,K,2),0.0)*CSS**(K-1)
08400      743      CONTINUE
08500      CNUM=CNUM+CMPLX(G(I,J,1,1),0.0)
08600      CDOM=CDOM+CMPLX(G(I,J,1,2),0.0)
08700      IF(CDOM.EQ.(0.0,0.0))GO TO 723
08800      C=CNUM/CDOM
08900      XR1(I,J,N+99+NTIMES)=REAL(C)
09000      XI1(I,J,N+99+NTIMES)=AIMAG(C)
09100      CONTINUE
770      CONTINUE
780      RETURN
09300      END
09400
09500      C      *****
09600      C      NO INTERACTIVE USAGE
09700      C      *****
09800      C      NO ADDITIONAL ROUTINES CALLED
09900      C      *****
10000      C      *****
10100      C      *****
10200      C      SUBROUTINE MAG
10300      C      *****
10400      C      TAKES COMPLEX VALUE OF XC INTO (XR1,XI1)
10500      C      *****
10600      C      IMPLICIT COMPLEX(C)
10700      C      COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,IO,MO
10800      C      *****
10900      C      USES COMMON/A32/,COMMON/A7/ ALSO
11000      C      *****
11100      C      COMMON/A32/XR1(2,2,201),XI1(2,2,201)
11200      C      COMMON/A7/XS(2,2),XC(2,2,6,2)
11300      C      *****
11400      C      DO 60 I=1,MO
11500      C      DO 60 J=1,MO
11600      C      DO 50 N=1,IEND
11700      C      OMEGA=OMEG0+(N-1)*DTOMEG
11800      C      SIGMA=0.0
11900      C      GO TO 30
12000      20      SIGMA=-DTOMEG

```

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END

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```

12100      TYPE 900,OMEGA,I,J,N
12200  900  FORMAT(X,'A POLE AT OMEGA = ',F7.2,2X,'FOR FUNCTION ('',I1,'',',
12300      I1,'',',I3,'')')
12400  30    CSS=CMPLX(SIGMA,OMEGA)
12500      CNUM=(0.0,0.0)
12600      CDOM=(0.0,0.0)
12700      DO 40 K=2,6
12800      CNUM=CNUM+CMPLX(XC(I,J,K,1),0.0)*CSS**(K-1)
12900      CDOM=CDOM+CMPLX(XC(I,J,K,2),0.0)*CSS**(K-1)
13000  40    CONTINUE
13100      CNUM=CNUM+CMPLX(XC(I,J,1,1),0.0)
13200      CDOM=CDOM+CMPLX(XC(I,J,1,2),0.0)
13300      IF(CDOM.EQ.(0.0,0.0)) GO TO 20
13400      C=CNUM/CDOM
13500      XR1(I,J,N)=REAL(C)
13600      XI1(I,J,N)=AIMAG(C)
13700  50    CONTINUE
13800  60    CONTINUE
13900      RETURN
14000      END
14100  C      *****
14200  C      NO INTERACTIVE USAGE
14300  C      *****
14400  C      NO ADDITIONAL ROUTINES CALLED
14500  C      *****
14600  C      *****
14700  C      *****
14800      SUBROUTINE MIMP
14900  C      *****
15000  C      TAKES MATRIX MULTP. OF XS=CMPLX(XR1,XI1)
15100  C      STORES THE RESULT IN Q
15200  C      *****
15300      IMPLICIT COMPLEX(C)
15400      COMMON/A2/D1,D2,D3,ISTART,IEND,IO,MO
15500  C      *****
15600  C      USES COMMON/A31/,COMMON/A32/,COMMON/A4/,COMMON/A7/ ALSO
15700  C      *****
15800      COMMON/A31/C1(2,2),C2(2,2),S1(2),S2(2)
15900      COMMON/A32/XR1(2,2,201),XI1(2,2,201)
16000      COMMON/A4/QR(2,2,201),QI(2,2,201)
16100      COMMON/A7/XS(2,2),XC(2,2,6,2)
16200  C      *****
16300      DO 40 N=1,IEND
16400      DO 10 I=1,MO
16500      DO 10 J=1,MO
16600      C1(I,J)=CMPLX(XR1(I,J,N),XI1(I,J,N))
16700  10    CONTINUE
16800      DO 30 I=1,MO
16900      DO 30 J=1,MO
17000      CX=(0.0,0.0)
17100      DO 20 K=1,MO
17200      CX=CX+XS(I,K)*C1(K,J)
17300  20    CONTINUE
17400      QR(I,J,N)=REAL(CX)
17500      QI(I,J,N)=AIMAG(CX)
17600  30    CONTINUE
17700  40    CONTINUE
17800      RETURN
17900      END
18000  C      *****

```

```

18100 C NO INTERACTIVE USAGE
18200 C *****
18300 C NO ADDITIONAL ROUTINES CALLED
18400 C *****
18500 C *****
18600 C *****
18700 C SUBROUTINE MCMP
18800 C *****
18900 C TAKES MATRIX MULTP. OF CMPLX(XR1,XI1)*CMPLX(XR2,XI2)
19000 C STORES RESULT IN Q
19100 C *****
19200 C IMPLICIT COMPLEX(C)
19300 C COMMON/A2/D1,D2,D3,ISTART,IEND,ID,MO
19400 C *****
19500 C USES COMMON/A31/,COMMON/A32/,COMMON/A33/,COMMON/A4/ ALSO
19600 C *****
19700 C COMMON/A31/C1(2,2),C2(2,2),S1(2),S2(2)
19800 C COMMON/A32/XR1(2,2,201),XI1(2,2,201)
19900 C COMMON/A33/XR2(2,2,201),XI2(2,2,201)
20000 C COMMON/A4/QR(2,2,201),OI(2,2,201)
20100 C *****
20200 C DO 50 N=1,IEND
20300 C DO 20 I=1,MO
20400 C DO 20 J=1,MO
20500 C C1(I,J)=CMPLX(XR1(I,J,N),XI1(I,J,N))
20600 C C2(I,J)=CMPLX(XR2(I,J,N),XI2(I,J,N))
20700 20 CONTINUE
20800 C CX=(0.0,0.0)
20900 C DO 30 K=1,MO
21000 C CX=CX+C1(I,K)*C2(K,J)
21100 30 CONTINUE
21200 C QR(I,J,N)=REAL(CX)
21300 C OI(I,J,N)=AIMAG(CX)
21400 40 CONTINUE
21500 50 CONTINUE
21600 C RETURN
21700 C END
21800 C *****
21900 C NO INTERACTIVE USAGE
22000 C *****
22100 C NO ADDITIONAL ROUTINES USED
22200 C *****
22300 C *****
22400 C *****
22500 C SUBROUTINE GERSH
22600 C *****
22700 C COMPUTES GERSHGORIN BANDS FOR CURRENT Q VALUES
22800 C PLACES RESULT IN COMMON/A5/ BANDS
22900 C *****
23000 C IMPLICIT COMPLEX(C)
23100 C COMMON/A2/D1,D2,D3,ISTART,IEND,ID,MO
23200 C *****
23300 C USES COMMON/A31/,COMMON/A4/,COMMON/A5/ ALSO
23400 C *****
23500 C COMMON/A31/C1(2,2),C2(2,2),S1(2),S2(2)
23600 C COMMON/A4/QR(2,2,201),OI(2,2,201)
23700 C COMMON/A5/BAND(2,2,201,2)
23800 C *****
23900 C DO 30 N=1,IEND
24000 C DO 10 I=1,MO

```



```

24100      DO 10 J=1,M0
24200      C2(I,J)=CHPLX(OR(I,J,N),QI(I,J,N))
24300      10 CONTINUE
24400      DO 30 I=1,M0
24500      DO 30 J=1,M0
24600      XR=0.0
24700      XC=0.0
24800      DO 20 K=1,M0
24900      IF(K.NE.J) XR=XR+CABS(C2(I,K))
25000      IF(K.NE.I) XC=XC+CABS(C2(K,J))
25100      20 CONTINUE
25200      BAND(I,J,N,1)=XR
25300      BAND(I,J,N,2)=XC
25400      30 CONTINUE
25500      RETURN
25600      END
25700      C *****
25800      C NO INTERACTIVE USAGE
25900      C *****
26000      C NO ADDITIONAL ROUTINES USED
26100      C *****
26200      C *****
26300      C *****
26400      SUBROUTINE OSTROW
26500      C *****
26600      C COMP. OSTROWSKI BANDS FROM CURRENT GERSH, BANDS,Q AND F VALUES
26700      C PLACES RESULT IN COMMON/A6/ OST
26800      C *****
26900      C IMPLICIT COMPLEX(C)
27000      COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,M0
27100      C *****
27200      C USES COMMON/A7/,COMMON/A4/,COMMON/A5/,COMMON/A6/ ALSO
27300      C *****
27400      COMMON/A4/OR(2,2,201),QI(2,2,201)
27500      COMMON/A5/BAND(2,2,201,2)
27600      COMMON/A6/OST(2,201,2)
27700      COMMON/A7/XS(2,2),XC(2,2,6,2)
27800      C *****
27900      DO 40 N=1,IEND
28000      XOR=0.0
28100      XOC=0.0
28200      DO 30 I=1,M0
28300      DO 20 J=1,M0
28400      IF(J.EQ.I) GO TO 20
28500      C=CHPLX((XS(J,J)+OR(J,J,N)),QI(J,J,N))
28600      H=CABS(C)
28700      XOR1=BAND(J,J,N,1)/H
28800      XOC1=BAND(J,J,N,2)/H
28900      IF(XOR1.GT.XOR) XOR=XOR1
29000      IF(XOC1.GT.XOC) XOC=XOC1
29100      20 CONTINUE
29200      OST(I,N,1)=XOR
29300      OST(I,N,2)=XOC
29400      30 CONTINUE
29500      40 CONTINUE
29600      RETURN
29700      END
29800      C *****
29900      C NO INTERACTIVE USAGE
30000      C *****

```

```

30100 C NO ADDITIONAL ROUTINES USED
30200 C *****
30300 C *****
30400 C *****
30500 SUBROUTINE MMIJ(I,J,XMIN,XMAX,YMIN,YMAX)
30600 C *****
30700 C SEARCHS (I,J) COMPOSIT NYQUIST DIAGRAM FOR (X,Y) MIN/MAX
30800 C *****
30900 COMMON/A2/D1,D2,D3,ISTART,IEND,NIDI,MO
31000 C *****
31100 C USES COMMON/A4,COMMON/A5/ ALSO
31200 C *****
31300 COMMON/A4/QR(2,2,201),QI(2,2,201)
31400 COMMON/A5/BAND(2,2,201,2)
31500 C *****
31600 DO 20 NL=1,IEND
31700 IRC=1
31800 IF(BAND(I,J,NL,1).GE.BAND(I,J,NL,2)) IRC=2
31900 XMIN1=QR(I,J,NL)=BAND(I,J,NL,IRC)
32000 XMAX1=QR(I,J,NL)+BAND(I,J,NL,IRC)
32100 IF(I.NE.J) XMIN1=QR(I,J,NL)
32200 IF(I.NE.J) XMAX1=QR(I,J,NL)
32300 IF(XMIN1.LT.XMIN) XMIN=XMIN1
32400 IF(XMAX1.GT.XMAX) XMAX=XMAX1
32500 YMIN1=QI(I,J,NL)=BAND(I,J,NL,IRC)
32600 YMAX1=QI(I,J,NL)+BAND(I,J,NL,IRC)
32700 IF(I.NE.J) YMIN1=QI(I,J,NL)
32800 IF(I.NE.J) YMAX1=QI(I,J,NL)
32900 IF(YMIN1.LT.YMIN) YMIN=YMIN1
33000 IF(YMAX1.GT.YMAX) YMAX=YMAX1
33100 20 CONTINUE
33200 RETURN
33300 END
33400 C *****
33500 C NO INTERACTIVE USAGE
33600 C *****
33700 C NO ADDITIONAL ROUTINES CALLED
33800 C *****
33900 C *****

```

Table A.2.2. GRAPH.FOR listing

```

00100 C *****
00200 C *****
00300 C SUBROUTINE PTNY1(I,J)
00400 C *****
00500 C PLOTS (I,J) NYQUIST DIAGRAM
00600 C *****
00700 C COMMON/A2/D1,D2,D3,ISTART,IEND,IO,MO
00800 C *****
00900 C USES COMMON/A4/ ALSO
01000 C *****
01100 C COMMON/A4/QR(2,2,201),QI(2,2,201)
01200 C *****
01300 C DIMENSION XR(202),XI(202)
01400 C NOTE : FOR PLOTTING MORE THAN 201 POINTS IN Q CHANGE
01500 C DIMENSIONS ACCORDINGLY
01600 C IMANY=IEND-ISTART+1
01700 C XR(1)=IMANY
01800 C XI(1)=IMANY
01900 C DO 10 N=1,IMANY
02000 C XR(N+1)=QR(I,J,N+ISTART-1)
02100 C XI(N+1)=QI(I,J,N+ISTART-1)
02200 10 CONTINUE
02300 C CALL CHECK(XR,XI)
02400 C CALL DSPLAY(XR,XI)
02500 C RETURN
02600 C END
02700 C *****
02800 C NO INTERACTIVE USAGE
02900 C *****
03000 C NO ADDITIONAL ROUTINES USED
03100 C *****
03200 C *****
03300 C *****
03400 C SUBROUTINE PTGR1(I,J,IRC)
03500 C *****
03600 C PLOTS GERSHGORIN BANDS OF (I,J) ENTRY OF Q
03700 C IRC = 1 FOR PLOTTING BANDS BASED ON ROWS
03800 C      2 FOR PLOTTING BANDS BASED ON COLUMNS
03900 C *****
04000 C COMMON/A2/DMEGO,DMEGND,DTOMEG,ISTART,IEND,NP,MO
04100 C *****
04200 C USES COMMON/A4/,COMMON/A5/
04300 C COMMON/A4/QR(2,2,201),QI(2,2,201)
04400 C COMMON/A5/BAND(2,2,201,2)
04500 C *****
04600 C DIMENSION X(102),Y(102)
04700 C NSTOP=(IEND-ISTART+1)/NP+1
04800 C X(1)=101
04900 C Y(1)=101
05000 C DO 20 N=1,NSTOP
05100 C NI=(N-1)*NP+ISTART
05200 C DO 10 K=1,101
05300 C X(K+1)=QR(I,J,NI)+BAND(I,J,NI,IRC)*COS((K-1)*.0628318)
05400 C Y(K+1)=QI(I,J,NI)+BAND(I,J,NI,IRC)*SIN((K-1)*.0628318)
05500 10 CONTINUE
05600 C CALL CPLOT(X,Y)
05700 20 CONTINUE
05800 C RETURN
05900 C END
06000 C *****

```



```

06100 C NO INTERACTIVE USAGE
06200 C *****
06300 C NO ADDITIONAL ROUTINES CALLED
06400 C *****
06500 C *****
06600 C *****
06700 C SUBROUTINE PTOST(I,IRC)
06800 C *****
06900 C PLOTS OSTROWSKI BANDS FOR H(I,I)
07000 C IRC = 1 FOR BANDS BASED ON ROWS
07100 C       = 2 FOR BANDS BASED ON COLUMNS
07200 C *****
07300 C COMMON/A2/OMEG0,OMEGND,OTOMEG,ISTART,IEND,NP,MO
07400 C *****
07500 C USES COMMON/A4/,COMMON/A6/ ALSO
07600 C *****
07700 C COMMON/A4/OR(2,2,201),OI(2,2,201)
07800 C COMMON/A6/OST(2,201,2)
07900 C *****
08000 C DIMENSION X(102),Y(102)
08100 C NSTOP=(IEND-ISTART+1)/NP+1
08200 C X(1)=101
08300 C Y(1)=101
08400 C DO 20 N=1,NSTOP
08500 C NI=(N-1)*NP+ISTART
08600 C DO 10 K=1,101
08700 C X(K+1)=OR(I,I,NI)+OST(I,NI,IRC)*COS((K-1)*.0628318)
08800 C Y(K+1)=OI(I,I,NI)+OST(I,NI,IRC)*SIN((K-1)*.0628318)
08900 10 CONTINUE
09000 C CALL CPLOT(X,Y)
09100 20 CONTINUE
09200 C RETURN
09300 C END
09400 C *****
09500 C NO INTERACTIVE USAGE
09600 C *****
09700 C NO ADDITIONAL ROUTINES USED
09800 C *****
09900 C *****
10000 C *****
10100 C SUBROUTINE PTBIG1
10200 C *****
10300 C PLOTS NYQUIST DIAGRAMS W/OSTROWSKI BANDS
10400 C USES MAXIMUM SCREEN SIZE
10500 C *****
10600 C NO COMMON BLOCK USAGE
10700 C *****
10800 901 FORMAT(I1,X,I1)
10900 10 CONTINUE
11000 C CALL ANMODE
11100 C CALL HOME
11200 C TYPE 900
11300 900 FORMAT(X,'TYPE 0, OR TYPE I, 1/2 FOR H(I,I) AND OST ROWS/COLUMNS')
11400 C ACCEPT 901,I,IRC
11500 C IF(I.EQ.0) GO TO 20
11600 C CALL BINITT
11700 C CALL NEWPAG
11800 C CALL SLIMX(75,1000)
11900 C CALL SLIMY(50,740)
12000 C CALL PTNY1(I,I)

```



```

12100      CALL PTOST1(I,IRC)
12200      GO TO 10
12300      CONTINUE
12400      CALL ANNODE
12500      CALL NEWPAG
12600      RETURN
12700      END
12800      C *****
12900      C INTERACTIVELY ACCEPTS:
13000      C      1) DIAGONAL ELEMENT
13100      C      2) BAND BASED ON ROW/COLUMN
13200      C *****
13300      C ADDITIONAL ROUTINES CALLED : PTNY1,PTOST1
13400      C *****
13500      C *****
13600      C *****
13700      C SUBROUTINE MAK4(IRC)
13800      C *****
13900      C PLOTS ALL NYQUIST DIAGRAMS FOR (2X2) SYSTEM ON SCREEN
14000      C *****
14100      C COMMON/A2/01,02,03,100,101,102,10
14200      C *****
14300      C NO ADDITIONAL COMMON USAGE
14400      C *****
14500      C XMIN=10000.
14600      C YMIN=10000.
14700      C XMAX=-10000.
14800      C YMAX=-10000.
14900      C DO 10 I=1,M0
15000      C DO 10 J=1,M0
15100      C ID=I
15200      C JD=J
15300      C CALL MMIJ(ID,JD,XMIN,XMAX,YMIN,YMAX)
15400      C CONTINUE
15500      C CALL OLIMX(XMIN,XMAX)
15600      C CALL OLIMY(YMIN,YMAX)
15700      C CALL ERASE
15800      C CALL PLACE(4)
15900      C CALL PTNY1(1,1)
16000      C IF(IRC.NE.0) CALL PTGR1(1,1,IRC)
16100      C CALL PLACE(5)
16200      C CALL PTNY1(1,2)
16300      C CALL PLACE(6)
16400      C CALL PTNY1(2,1)
16500      C CALL PLACE(7)
16600      C CALL PTNY1(2,2)
16700      C IF(IRC.NE.0) CALL PTGR1(2,2,IRC)
16800      C RETURN
16900      C END
17000      C *****
17100      C NO INTERACTIVE USAGE
17200      C *****
17300      C ADDITIONAL ROUTINES CALLED : MMIJ,PTNY1,PTGR1
17400      C *****
17500      C *****
17600      C *****
17700      C SUBROUTINE DRW
17800      C *****
17900      C PLOT ROUTING ROUTINE
18000      C *****

```

```

18100 C      NO COMMON BLOCK USAGE
18200 C      *****
18300 10     CALL ANMODE
18400      CALL HOME
18500      CALL NEWLIN
18600      CALL NEWLIN
18700      CALL NEWLIN
18800      TYPE 900
18900 900    FORMAT(5X,'PLOT DIAGRAMS? TYPE: 0 FOR NO,1 FOR YES')
19000      ACCEPT *,IDUM
19100      IF(IDUM.EQ.0) GO TO 100
19200      ENTRY ORW1
19300      CALL ANMODE
19400      TYPE 901
19500 901    FORMAT(X,'ENTER -0,0- FOR 4,OR -I,J- FOR Q(I,J)')
19600      ACCEPT *,I,J
19700      IF(I.EQ.0.AND.J.EQ.0) GO TO 20
19800      XMIN=100000.
19900      YMIN=100000.
20000      XMAX=-100000.
20100      YMAX=-100000.
20200      CALL MMIJ(I,J,XMIN,XMAX,YMIN,YMAX)
20300      CALL DLIMX(XMIN,XMAX)
20400      CALL DLIMY(YMIN,YMAX)
20500      CALL ERASE
20600      CALL PLACE(1)
20700      CALL PTNY1(I,J)
20800 20     CONTINUE
20900      CALL HOME
21000      CALL NEWLIN
21100      TYPE 902
21200 902    FORMAT(5X,'PLOT BANDS? TYPE: 0 FOR NO,1 FOR ROWS,2 FOR COLUMNS')
21300      ACCEPT *,IRC
21400      IF(IRC.EQ.0.AND.I.NE.0) GO TO 10
21500      IF(I.EQ.0.AND.J.EQ.0) CALL MAK4(IRC)
21600      IF(I.EQ.0.AND.J.EQ.0) GO TO 10
21700      CALL PTGR1(I,J,IRC)
21800      GO TO 10
21900 100    CONTINUE
22000      RETURN
22100      END
22200 C      *****
22300 C      INTERACTIVELY ACCEPTS :
22400 C      1) PLOT YES/NO
22500 C      2) PLOT 4 DIAGRMS/I,J ELEMENT
22600 C      3) PLOT GERSH.BANDS NO/ROWS/COLUMNS
22700 C      *****
22800 C      ADITIONAL ROUTINES USED : MMIJ,PTNY1,PTGR1,MAK4
22900 C      *****
23000 C      *****

```

- 4) Subroutine QOUT outputs the numerical value of the open loop system matrix $\hat{Q}(s)$ to the terminal screen or lineprinter.
- 5) Subroutine QDSKWR interactively stores the open loop system $\hat{Q}(s)$ numerics on computer disk for the next design run.
- 6) Subroutine QDSKRD automatically reads the open loop system $\hat{Q}(s)$ numerics from disk so they do not have to be recalculated each time the program is run.
- 7) Subroutine OUT interactively outputs system matrices and programming variable to the terminal screen or lineprinter.
- 8) Subroutine FINISH notifies the user he is stopping, causes system outputs and disk storage to possibly be performed, and terminates program execution.

A listing of this program is given in Table A.2.3.

- IV. AUTOMP.FOR contains the main program used in investigating the autopilot program.
- V. OLISP.FOR contains the open loop constant input compensator design package.
- VI. IFDCP.FOR contains the open loop frequency dependent compensator design package.
- VII. CLSP.FOR contains the closed loop feedback design package.
- VIII. DATA12.FOR contains the subroutine defining the system and programming data for the autopilot problem.
- IX. QINV2.FOR contains the subroutine which takes the point by point inverse of the autopilot plant data.

See Table A.2.4 for a program listing.

Table A.2.3. AUX.FOR Listing

```

00100 C *****
00200 C *****
00300 SUBROUTINE MODIA2
00400 C *****
00500 C MODIFIES INTEGER VALUES IN COMMON/A2/
00600 C *****
00700 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
00800 C *****
00900 C NO OTHER COMMON BLOCK USAGE
01000 C *****
01100 TYPE 900
01200 900 FORMAT(X,'ENTER ISTART,IEND,NP,OR -,- AND RETURN')
01300 ACCEPT *,I1,I2,I3
01400 IF(I1.NE.0) ISTART=I1
01500 IF(I2.NE.0) IEND=I2
01600 IF(I3.NE.0) NP=I3
01700 TYPE *,ISTART,IEND,NP
01800 RETURN
01900 END
02000 C *****
02100 C INTERACTIVE USAGE
02200 C 1) ENTER VALUES OR COMMA
02300 C *****
02400 C NO OTHER ROUTINES USES
02500 C *****
02600 C *****
02700 C *****
02800 SUBROUTINE MODXS
02900 C *****
03000 C MODIFIES AND OUTPUTS MATRIX XS
03100 C *****
03200 COMMON/A2/D1,D2,D3,I00,I01,I02,MO
03300 C *****
03400 C USES COMMON/A7/ ALSO
03500 C *****
03600 COMMON/A7/XS(2,2),XC(2,2,6,2)
03700 C *****
03800 CALL ANMODE
03900 10 TYPE 901
04000 901 FORMAT(X,'MODIFY XS? TYPE: 0 FOR NO,1 FOR YES')
04100 ACCEPT *,IDUM
04200 IF(IDUM.EQ.0) GO TO 20
04300 TYPE 904
04400 904 FORMAT(X,'TYPE: I,J,X IN XS(I,J)=X')
04500 ACCEPT *,I,J,X
04600 XS(I,J)=X
04700 GO TO 10
04800 20 TYPE 902
04900 902 FORMAT(X,'OUTPUT XS? TYPE: 0 FOR NO,1 FOR TTY,2 FOR LP')
05000 ACCEPT *,IDUM
05100 IF(IDUM.EQ.0) GO TO 40
05200 DO 30 I=1,MO
05300 DO 30 J=1,MO
05400 IF(IDUM.EQ.1) TYPE 903,I,J,XS(I,J)
05500 IF(IDUM.EQ.2) PRINT 903,I,J,XS(I,J)
05600 903 FORMAT(X,'XS(',I1,',',J1,',') = ',F7,2)
05700 30 CONTINUE
05800 40 CONTINUE
05900 RETURN
06000 END

```

```

06100 C *****
06200 C INTERACTIVE USAGE:
06300 C 1) MODIFY XS YES/NO
06400 C 2) ENTER VALUES
06500 C 3) OUTPUT XS NO/TTY/LP
06600 C *****
06700 C NO ADDITIONAL ROUTINES USED
06800 C *****
06900 C *****
07000 C *****
07100 C SUBROUTINE MOOXC
07200 C *****
07300 C MODIFIES AND OUTPUTS MATRIX XC
07400 C *****
07500 C COMMON/A2/D1,D2,D3,ID0,ID1,ID2,M0
07600 C *****
07700 C USES COMMON/A7/ ALSO
07800 C *****
07900 C COMMON/A7/XS(2,2),XC(2,2,6,2)
08000 C *****
08100 C CALL ANMODE
08200 10 TYPE 901
08300 701 FORMAT(X,'MODIFY XC? TYPE: 0 FOR NO, 1 FOR YES')
08400 ACCEPT *,IDUM
08500 IF(IDUM.EQ.0) GO TO 20
08600 TYPE 902
08700 902 FORMAT(X,'I,J,N,X,1/2 IN XC(I,J)=X*(S)**N, NUM/COM')
08800 ACCEPT *,I,J,N,X,ND
08900 XC(I,J,N+1,ND)=X
09000 GO TO 10
09100 20 CONTINUE
09200 TYPE 903
09300 903 FORMAT(X,'OUTPUT XC? TYPE: 0 FOR NO, 1 FOR TTY, 2 FOR LP')
09400 ACCEPT *,IDUM
09500 IF(IDUM.EQ.0) GO TO 60
09600 IF(IDUM.EQ.2) GO TO 40
09700 DO 30 I=1,M0
09800 DO 30 J=1,M0
09900 TYPE 904,XC(I,J,3,1),XC(I,J,2,1),XC(I,J,1,1)
10000 904 FORMAT(11X,F7,2,3X,F7,2,3X,F7,2)
10100 TYPE 905,I,J
10200 905 FORMAT(X,'XC('',I1,'',',',I1,'') = ',29(''-'))
10300 TYPE 904,XC(I,J,3,2),XC(I,J,2,2),XC(I,J,1,2)
10400 TYPE 906
10500 906 FORMAT(X,' ')
10600 30 CONTINUE
10700 40 IF(IDUM.NE.2)GO TO 60
10800 DO 50 I=1,M0
10900 DO 50 J=1,M0
11000 PRINT 904,XC(I,J,3,1),XC(I,J,2,1),XC(I,J,1,1)
11100 PRINT 905,I,J
11200 PRINT 904,XC(I,J,3,2),XC(I,J,2,2),XC(I,J,1,2)
11300 PRINT 906
11400 50 CONTINUE
11500 60 CONTINUE
11600 RETURN
11700 END
11800 C *****
11900 C INTERACTIVE USAGE
12000 C 1) MODIFY XC YES/NO

```

```

12100 C          2) ENTER VALUES
12200 C          3) OUTPUT XC NO/TTY/LP
12300 C *****
12400 C NO ADDITIONAL ROUTINES USED
12500 C *****
12600 C *****
12700 C *****
12800 C SUBROUTINE QOUT(IOUT)
12900 C *****
13000 C OUTPUTS CURRENT Q VALUES IN COMMON/A4/
13100 C IOUT = 1 FOR TTY
13200 C       = 2 FOR LP
13300 C *****
13400 C IMPLICIT COMPLEX(C)
13500 C COMMON/A2/OMEGA,OMEGNO,OTOMEG,ISTART,IEND,NP,MO
13600 C *****
13700 C USES COMMON/A4/ ALSO
13800 C *****
13900 C COMMON/A4/QR(2,2,201),QI(2,2,201)
14000 C *****
14100 C DO 20 I=1,MO
14200 C DO 20 J=1,MO
14300 C DO 10 N=ISTART,IEND
14400 C IF(IOUT.EQ.1) TYPE *,I,J,N,QR(I,J,N),QI(I,J,N)
14500 C IF(IOUT.EQ.2) PRINT *,I,J,N,QR(I,J,N),QI(I,J,N)
14600 10 CONTINUE
14700 20 CONTINUE
14800 RETURN
14900 END
15000 C *****
15100 C NO INTERACTIVE USAGE
15200 C *****
15300 C NO ADDITIONAL ROUTINES USED
15400 C *****
15500 C *****
15600 C *****
15700 C SUBROUTINE QDSKWR
15800 C *****
15900 C WRITES Q DATA TO DISK FILE Q1.DAT
16000 C *****
16100 C COMMON/A2/D1,D2,D3,ISTART,IEND,ID1,MO
16200 C *****
16300 C USES COMMON/A4/ ALSO
16400 C *****
16500 C COMMON/A4/QR(2,2,201),QI(2,2,201)
16600 C *****
16700 C TYPE 888
16800 888 FORMAT(X,'STORE Q VALUES? TYPE 0 FOR NO, 1 FOR YES')
16900 ACCEPT *,ID0
17000 IF(ID0.EQ.0) GO TO 40
17100 IDUM=1
17200 OPEN(UNIT=1,ACCESS='SEQUENT',FILE='Q1.DAT')
17300 WRITE(1,900) IDUM
17400 900 FORMAT(X,I1)
17500 DO 30 N=1,IEND
17600 DO 30 I=1,MO
17700 DO 30 J=1,MO
17800 WRITE(1,901) I,J,N,QR(I,J,N),QI(I,J,N)
17900 901 FORMAT(X,I1,X,I1,X,I3,X,E14.7,X,E14.7)
18000 30 CONTINUE

```



```

18100      CLOSE(UNIT=1,ACCESS='SEQOUT',FILE='Q1,DAT')
18200      40      CONTINUE
18300      RETURN
18400      END
18500      C      *****
18600      C      INTERACTIVE USAGE:
18700      C      1) STORE Q VALUES NO/YES
18800      C      *****
18900      C      NO ADDITIONAL ROUTINES USED
19000      C      *****
19100      C      *****
19200      C      *****
19300      C      SUBROUTINE QDSKRD(IDUM)
19400      C      *****
19500      C      READS Q DATA FROM DISK FILE Q1,DAT
19600      C      IDUM = 0 ON RETURN FOR NO PREVIOUSLY STORED DATA
19700      C      = 1 ON RETURN IF Q VALUES ARE READ FROM DISK
19800      C      *****
19900      C      COMMON/A2/D1,D2,D3,ISTART,IEND,ID1,MO
20000      C      *****
20100      C      USES COMMON/A4/ ALSO
20200      C      *****
20300      C      COMMON/A4/QR(2,2,201),QI(2,2,201)
20400      C      *****
20500      900      FORMAT(X,I1)
20600      901      FORMAT(X,I1,X,I1,X,I3,X,E14.7,X,E14.7)
20700      OPEN(UNIT=1,ACCESS='SEQIN',FILE='Q1,DAT')
20800      READ(1,900) IDUM
20900      IF(IDUM.EQ.0) GO TO 30
21000      10      READ(1,901,END=20) I,J,N,SR,SI
21100      QR(I,J,N)=SR
21200      QI(I,J,N)=SI
21300      GO TO 10
21400      20      CONTINUE
21500      IEND=N
21600      30      CONTINUE
21700      CLOSE(UNIT=1,ACCESS='SEQIN',FILE='Q1,DAT')
21800      RETURN
21900      END
22000      C      *****
22100      C      NO INTERACTIVE USAGE
22200      C      *****
22300      C      NO ADDITIONAL ROUTINES USED
22400      C      *****
22500      C      *****
22600      C      *****
22700      C      SUBROUTINE OUT
22800      C      *****
22900      C      OUTPUTS COMMON/A1/ VALUES
23000      C      *****
23100      C      COMMON/A2/OMEG0,OMEGND,OTOMEG,ISTART,IEND,NP,MO
23200      C      *****
23300      C      USES COMMON/A1/ ALSO
23400      C      *****
23500      C      COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XLB(2,2,3,2),
23600      C      1 XKA(2,2),XLA(2,2),F(2,2)
23700      C      *****
23800      900      FORMAT(X,'KA(',I1,',',I1,') = ',F7.2,3X,'LA(',I1,',',I1,
23900      C      ') = ',F7.2,3X,'F(',I1,',',I1,') = ',F7.2)
24000      901      FORMAT(10X,F7.2,3X,F7.2,3X,F7.2,3X,F7.2,3X,F7.2)

```

```

24100 902 FORMAT(X,'G(',I1,',',I1,') = ',50(''))
24200 903 FORMAT(11X,F7.2,3X,F7.2,3X,F7.2)
24300 904 FORMAT(X,'KH(',I1,',',I1,') = ',32(''))
24400 905 FORMAT(X,'LB(',I1,',',I1,') = ',32(''))
24500 906 FORMAT(X)
24600 907 FORMAT(X,'OMEG = ',F7.2,X,'OMEGND = ',F7.2,X,'DTOMEG = ',
24700 ,F7.2,X,'IEND = ',I3,X,'NP = ',I3)
24800 908 FORMAT(X,'THE ORDER OF THE SYSTEM IS = ',I1)
24900 909 FORMAT(X,'FOR OUTPUT TYPE:')
25000 910 FORMAT(I,' 0 FOR NONE,1 FOR TTY,2 FOR LP')
25100 911 FORMAT(X,'TYPE: 0 TO CONTINUE, TYPE: 1 TO DO AGAIN')
25200 912 FORMAT(I1)
25300 913 FORMAT(X,'FOR SYSTEM OUTPUT TYPE 0, FOR 0 OUTPUT TYPE 1')
25400 5
25500 TYPE 909
25600 TYPE 910
25700 ACCEPT *,I1
25800 IF(I1.EQ.0) RETURN
25900 TYPE 906
26000 TYPE 913
26100 ACCEPT *,I2
26200 IF(I2.EQ.1) CALL QOUT(I1)
26300 IF(I2.EQ.1) GO TO 90
26400 I1=I1-1
26500 IF(I1.EQ.1) GO TO 100
26600 DO 10 I=1,M0
26700 DO 10 J=1,M0
26800 TYPE 906
26900 TYPE 901,G(I,J,5,1),G(I,J,4,1),G(I,J,3,1),G(I,J,2,1),G(I,J,1,1)
27000 TYPE 902,I,J
27100 TYPE 901,G(I,J,5,2),G(I,J,4,2),G(I,J,3,2),G(I,J,2,2),G(I,J,1,2)
27200 TYPE 906
27300 10 CONTINUE
27400 DO 20 I=1,M0
27500 DO 20 J=1,M0
27600 TYPE 906
27700 TYPE 900,I,J,XKA(I,J),I,J,XLA(I,J),I,J,F(I,J)
27800 20 CONTINUE
27900 DO 30 I=1,M0
28000 DO 30 J=1,M0
28100 TYPE 906
28200 TYPE 903,XKB(I,J,3,1),XKB(I,J,2,1),XKB(I,J,1,1)
28300 TYPE 904,I,J
28400 TYPE 903,XKB(I,J,3,2),XKB(I,J,2,2),XKB(I,J,1,2)
28500 TYPE 906
28600 30 CONTINUE
28700 DO 40 I=1,M0
28800 DO 40 J=1,M0
28900 TYPE 906
29000 TYPE 903,XLB(I,J,3,1),XLB(I,J,2,1),XLB(I,J,1,1)
29100 TYPE 905,I,J
29200 TYPE 903,XLB(I,J,3,2),XLB(I,J,2,2),XLB(I,J,1,2)
29300 TYPE 906
29400 40 CONTINUE
29500 45 TYPE 907,OMEG,OMEGND,DTOMEG,IEND,NP
29600 TYPE 908,M0
29700 90 CONTINUE
29800 TYPE 911
29900 ACCEPT 912,IDUM
30000 IF(IDUM.EQ.1) GO TO 5
IF(I1.EQ.0) RETURN

```

```

30100 100 CONTINUE
30200 DO 110 I=1,M0
30300 DO 110 J=1,M0
30400 PRINT 906
30500 PRINT 901,G(I,J,5,1),G(I,J,4,1),G(I,J,3,1),G(I,J,2,1),G(I,J,1,1)
30600 PRINT 902,I,J
30700 PRINT 901,G(I,J,5,2),G(I,J,4,2),G(I,J,3,2),G(I,J,2,2),G(I,J,1,2)
30800 PRINT 906
30900 110 CONTINUE
31000 DO 120 I=1,M0
31100 DO 120 J=1,M0
31200 PRINT 906
31300 PRINT 900,I,J,XKA(I,J),I,J,XLA(I,J),I,J,F(I,J)
31400 120 CONTINUE
31500 DO 130 I=1,M0
31600 DO 130 J=1,M0
31700 PRINT 906
31800 PRINT 903,XKB(I,J,3,1),XKB(I,J,2,1),XKB(I,J,1,1)
31900 PRINT 904,I,J
32000 PRINT 903,XKB(I,J,3,2),XKB(I,J,2,2),XKB(I,J,1,2)
32100 PRINT 906
32200 130 CONTINUE
32300 DO 140 I=1,M0
32400 DO 140 J=1,M0
32500 PRINT 906
32600 PRINT 903,XLB(I,J,3,1),XLB(I,J,2,1),XLB(I,J,1,1)
32700 PRINT 905,I,J
32800 PRINT 903,XLB(I,J,3,2),XLB(I,J,2,2),XLB(I,J,1,2)
32900 PRINT 906
33000 140 CONTINUE
33100 145 PRINT 907,OMEG0,OMEGND,DTOMEG,IEND,NP
33200 PRINT 908,M0
33300 TYPE 911
33400 ACCEPT *,IDUM
33500 IF(IDUM,EQ.1) GO TO 5
33600 RETURN
33700 END
33800 C *****
33900 C INTERACTIVE USAGE
34000 C 1) OUTPUT DESIRED NONE/TTY/LP
34100 C 2) OUTPUT SYSTEM/Q ?
34200 C 3) DO AGAIN NO/YES
34300 C *****
34400 C NO ADDITIONAL ROUTINES USED
34500 C *****
34600 C *****
34700 C *****
34800 C SUBROUTINE FINISH
34900 C *****
35000 C STOPS EXECUTION OF THE MAIN PROGRAM
35100 C *****
35200 C NO COMMON BLOCK USAGE
35300 C *****
35400 C CALL ANMODE
35500 C TYPE 940
35600 940 FORMAT(X,'YOU ARE DONE!BEST PRINT OUT SYSTEM!!!')
35700 C CALL OUT
35800 C CALL NEWPAG
35900 C CALL QDSKWR
36000 C CALL FINITT(0,700)

```



```
36100      STOP
36200      RETURN
36300      END
36400      C *****
36500      C NO INTERACTIVE USAGE
36600      C *****
36700      C ADDITIONAL ROUTINES CALLED:OUT,QDSKWR
36800      C *****
```

Table A.2.4. Autopilot Design Program

```

00100 C
00200 C
00300 C *****
00400 C
00500 C THIS IS THE MAIN PROGRAM FOR THE AUTOPILOT DESIGN
00600 C
00700 C *****
00800 C IMPLICIT COMPLEX(C)
00900 C COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XLB(2,2,3,2),
01000 C 1 XKA(2,2),XLA(2,2),F(2,2)
01100 C COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
01200 C COMMON/A31/C1(2,2),C2(2,2),S1(2),S2(2)
01300 C COMMON/A32/XR1(2,2,201),XI1(2,2,201)
01400 C COMMON/A33/XR2(2,2,201),XI2(2,2,201)
01500 C COMMON/A4/QR(2,2,201),QI(2,2,201)
01600 C COMMON/A5/BAND(2,2,201,2)
01700 C COMMON/A6/OST(2,201,2)
01800 C COMMON/A7/XS(2,2),XC(2,2,6,2)
01900 C *****
02000 C CALL INITT(120)
02100 C CALL BINITT
02200 C CALL DATA12
02300 C IEND=(OMEGND-OMEG0)/DTOMEG+1
02400 C IF(IEND.GT.201) IEND=201
02500 C CALL QOSKRD(IDUM)
02600 C IF(IDUM.EQ.1) TYPE 888
02700 888 FORMAT(X,'0 VALUES WERE OBTAINED FROM DISK')
02800 C IF(IDUM.EQ.1) GO TO 10
02900 C DO 5 I=1,MO
03000 C DO 5 J=1,MO
03100 C XS(I,J)=0.0
03200 C IF(I.EQ.J) XS(I,J)=1.0
03300 5 CONTINUE
03400 C CALL MAP(0)
03500 C CALL MIMP
03600 C CALL QINV
03700 10 CONTINUE
03800 C TYPE 900
03900 900 FORMAT(X,'TYPE: 0 TO STOP')
04000 C TYPE 901
04100 901 FORMAT(X,'TYPE: 1 FOR OPEN LOOP PACKAGE')
04200 C TYPE 902
04300 902 FORMAT(X,'TYPE: 2 FOR INPUT COMPENSATOR PACKAGE')
04400 C TYPE 903
04500 903 FORMAT(X,'TYPE: 3 FOR CLOSED LOOP PACKAGE')
04600 C TYPE 904
04700 904 FORMAT(X,'TYPE: 4 FOR OUTPUT')
04800 C TYPE 905
04900 905 FORMAT(X,'TYPE: 5 TO MODIFY INTEGER COMMON/A2/ VALUES')
05000 C ACCEPT *,IDUM
05100 C IF(IDUM.EQ.0) CALL FINISH
05200 C IF(IDUM.EQ.1) CALL OLISP
05300 C IF(IDUM.EQ.2) CALL IFDCP
05400 C IF(IDUM.EQ.3) CALL CLSP
05500 C IF(IDUM.EQ.4) CALL OUT
05600 C IF(IDUM.EQ.5) CALL MODIA2
05700 C GO TO 10
05800 C STOP
05900 C END
06000 C *****

```

```

00100 C *****
00200 SUBROUTINE OLISP
00300 C *****
00400 IMPLICIT COMPLEX(C)
00500 COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XLB(2,2,3,2),
00600 1 XKA(2,2),XLA(2,2),F(2,2)
00700 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
00800 COMMON/A3/C1(2,2),C2(2,2),S1(2),S2(2)
00900 COMMON/A32/XR1(2,2,201),XI1(2,2,201)
01000 COMMON/A33/XR2(2,2,201),XI2(2,2,201)
01100 COMMON/A4/OR(2,2,201),OI(2,2,201)
01200 COMMON/A5/RAND(2,2,201,2)
01300 COMMON/A6/DST(2,201,2)
01400 COMMON/A7/XS(2,2),XC(2,2,6,2)
01500 C *****
01600 DO 110 I=1,MO
01700 DO 110 J=1,MO
01800 XS(I,J)=XKA(I,J)
01900 DO 100 N=1,IEND
02000 XR1(I,J,N)=OR(I,J,N)
02100 XI1(I,J,N)=OI(I,J,N)
02200 100 CONTINUE
02300 110 CONTINUE
02400 CALL GERSH
02500 TYPE 910
02600 910 FORMAT(X,'DEALING W/ OPEN LOOP INPUT CONST. COMP')
02700 120 CONTINUE
02800 TYPE 900
02900 900 FORMAT(X,'TYPE: 0 TO RETURN, 1 TO PLOT, 2 TO MODIFY')
03000 ACCEPT *,IDUM
03100 CALL NEWPAG
03200 IF(IDUM,EQ,0) GO TO 140
03300 IF(IDUM,EQ,1) GO TO 130
03400 CALL MODXS
03500 GO TO 120
03600 130 CONTINUE
03700 CALL MIMP
03800 CALL GERSH
03900 CALL DRW1
04000 GO TO 120
04100 140 CONTINUE
04200 DO 150 I=1,MO
04300 DO 150 J=1,MO
04400 XKA(I,J)=XS(I,J)
04500 150 CONTINUE
04600 199 CONTINUE
04700 RETURN
04800 END

```



```

00100 C *****
00200 SUBROUTINE IFDCP
00300 C *****
00400 IMPLICIT COMPLEX(C)
00500 COMMON/A1/G(2,2,6,2),XK8(2,2,3,2),XL8(2,2,3,2),
00600 1 XKA(2,2),XLA(2,2),F(2,2)
00700 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
00800 COMMON/A3/C1(2,2),C2(2,2),S1(2),S2(2)
00900 COMMON/A32/XR1(2,2,201),XI1(2,2,201)
01000 COMMON/A33/XR2(2,2,201),XI2(2,2,201)
01100 COMMON/A4/OR(2,2,201),OI(2,2,201)
01200 COMMON/A5/RAND(2,2,201,2)
01300 COMMON/A6/OST(2,201,2)
01400 COMMON/A7/XS(2,2),XC(2,2,6,2)
01500 C *****
01600 DO 250 I=1,MO
01700 DO 250 J=1,MO
01800 DO 210 K=1,IEND
01900 XR2(I,J,K)=OR(I,J,K)
02000 XI2(I,J,K)=OI(I,J,K)
02100 210 CONTINUE
02200 DO 240 L=1,2
02300 DO 220 K=1,3
02400 XC(I,J,K,L)=XK8(I,J,K,L)
02500 220 CONTINUE
02600 DO 230 K=4,6
02700 XC(I,J,K,L)=0,0
02800 230 CONTINUE
02900 240 CONTINUE
03000 250 CONTINUE
03100 CALL GERSH
03200 TYPE 920
03300 920 FORMAT(X,'DEALING W/ FREQ,DEPNOT, INPUT COMP')
03400 260 CONTINUE
03500 TYPE 900
03600 900 FORMAT(X,'TYPE: 0 TO RETURN, 1 TO PLOT, 2 TO MODIFY')
03700 ACCEPT *,IDUM
03800 CALL NEWPAG
03900 IF(IDUM.EQ.0) GO TO 280
04000 IF(IDUM.EQ.1) GO TO 270
04100 CALL MOUXC
04200 GO TO 260
04300 270 CONTINUE
04400 CALL MAG
04500 CALL MCMP
04600 CALL GERSH
04700 CALL DRW1
04800 GO TO 260
04900 280 CONTINUE
05000 DO 290 I=1,MO
05100 DO 290 J=1,MO
05200 DO 290 K=1,3
05300 DO 290 L=1,2
05400 XK8(I,J,K,L)=XC(I,J,K,L)
05500 290 CONTINUE
05600 299 CONTINUE
05700 RETURN
05800 END
05900 C *****

```

```

00100 C *****
00200 SUBROUTINE CLSP
00300 C *****
00400 IMPLICIT COMPLEX(C)
00500 COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XLB(2,2,3,2),
00600 1 XKA(2,2),XLA(2,2),F(2,2)
00700 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
00800 COMMON/A31/C1(2,2),C2(2,2),S1(2),S2(2)
00900 COMMON/A32/XR1(2,2,201),X11(2,2,201)
01000 COMMON/A33/XR2(2,2,201),X12(2,2,201)
01100 COMMON/A4/OR(2,2,201),OI(2,2,201)
01200 COMMON/A5/BAND(2,2,201,2)
01300 COMMON/A6/OST(2,201,2)
01400 COMMON/A7/XS(2,2),XC(2,2,6,2)
01500 C *****
01600 DO 310 I=1,MO
01700 DO 310 J=1,MO
01800 XS(I,J)=F(I,J)
01900 310 CONTINUE
02000 CALL GERSH
02100 CALL OSTROW
02200 TYPE 930
02300 930 FORMAT(X,'DEALING W/ CLOSED LOOP SYSTEM')
02400 320 CONTINUE
02500 TYPE 900
02600 900 FORMAT(X,'TYPE: 0 TO RETURN, 1 TO PLOT, 2 TO MODIFY')
02700 ACCEPT *, IDUM
02800 CALL NEWPAG
02900 IF(IDUM.EQ.0) GO TO 340
03000 IF(IDUM.EQ.1) GO TO 330
03100 CALL MODXS
03200 GO TO 320
03300 330 CONTINUE
03400 CALL OSTROW
03500 CALL PTBIG1
03600 GO TO 320
03700 340 CONTINUE
03800 DO 350 I=1,MO
03900 DO 350 J=1,MO
04000 F(I,J)=XS(I,J)
04100 350 CONTINUE
04200 399 CONTINUE
04300 RETURN
04400 END
04500 C *****

```

```

00100 C *****
00200 SUBROUTINE DATA12
00300 C *****
00400 COMMON/A1/G(2,2,6,2),XKB(2,2,3,2),XL8(2,2,3,2),
00500 1 XKA(2,2),XLA(2,2),F(2,2)
00600 COMMON/A2/OMEG0,OMEGND,DTOMEG,ISTART,IEND,NP,MO
00700 C SETS COMMON BLOCK /A2/ VALUES
00800 OMEG0=0.0
00900 UMEGND=0.16
01000 DTOMEG=0.0008
01100 ISTART=1
01200 IEND=201
01300 NP=10
01400 MO=2
01500 C SETS ALL DEMON. OF G(S)=CHAR. POLYNOMIAL
01600 DO 10 I=1,MO
01700 DO 10 J=1,MO
01800 G(I,J,1,2)=0.0
01900 G(I,J,2,2)=-.00166
02000 G(I,J,3,2)=.365
02100 G(I,J,4,2)=.364
02200 G(I,J,5,2)=.029
02300 G(I,J,6,2)=1.0
02400 10 CONTINUE
02500 C SETS INDIVIDUAL NUMERATOR ENTRIES OF G(S)
02600 G(1,1,1,1)=-.0195
02700 G(1,1,2,1)=-.00893
02800 G(1,1,3,1)=-.3121
02900 G(1,1,4,1)=-.3807
03000 G(1,1,5,1)=0.0
03100 G(1,1,6,1)=0.0
03200 G(1,2,1,1)=0.0
03300 G(1,2,2,1)=-.447
03400 G(1,2,3,1)=-.0503
03500 G(1,2,4,1)=-.0404
03600 G(1,2,5,1)=0.0
03700 G(1,2,6,1)=0.0
03800 G(2,1,1,1)=.022635
03900 G(2,1,2,1)=-.000588
04000 G(2,1,3,1)=-.0178
04100 G(2,1,4,1)=.06715
04200 G(2,1,5,1)=0.0
04300 G(2,1,6,1)=0.0
04400 G(2,2,1,1)=0.0
04500 G(2,2,2,1)=.602
04600 G(2,2,3,1)=.06237
04700 G(2,2,4,1)=1.587
04800 G(2,2,5,1)=0.0
04900 G(2,2,6,1)=0.0
05000 C SETS XKA=1
05100 XKA(1,1)=1.0
05200 XKA(1,2)=0.0
05300 XKA(2,1)=0.0
05400 XKA(2,2)=1.0
05500 RETURN
05600 END
05700 C *****

```



```

00100 C *****
00200 SUBROUTINE QINV
00300 C *****
00400 C TAKES 'LONG HAND' INVERSE OF Q
00500 C FOR A 2 X 2 SYSTEM ONLY
00600 C *****
00700 IMPLICIT COMPLEX(C)
00800 COMMON/A2/OMEG0,OMEGNO,DTOMEG,ISTART,IEND,NP,MO
00900 COMMON/A3/C1(2,2),C2(2,2),S1(2),S2(2)
01000 COMMON/A4/QR(2,2,201),QI(2,2,201)
01100 C *****
01200 DO 30 N=1,IEND
01300 DO 10 I=1,2
01400 DO 10 J=1,2
01500 C1(I,J)=CMPLX(QR(I,J,N),QI(I,J,N))
01600 10 CONTINUE
01700 XREAL=QR(1,1,N)*QR(2,2,N)+QI(1,2,N)*QI(2,1,N)
01800 1 -QI(1,1,N)*QI(2,2,N)-QR(1,2,N)*QR(2,1,N)
01900 XIMAG=QI(1,1,N)*QR(2,2,N)+QR(1,1,N)*QI(2,2,N)
02000 1 -QR(1,2,N)*QI(2,1,N)-QI(1,2,N)*QR(2,1,N)
02100 CCC=CMPLX(XREAL,XIMAG)
02200 IF(CCC.EQ.CMPLX(0.0,0.0)) TYPE 900,N
02300 900 FORMAT(X,'Q SINGULAR UPON INVERSION # ',I3)
02400 IF(CCC.EQ.CMPLX(0.0,0.0)) GO TO 15
02500 C2(1,1)=C1(2,2)/CCC
02600 C2(1,2)=-C1(1,2)/CCC
02700 C2(2,1)=-C1(2,1)/CCC
02800 C2(2,2)=C1(1,1)/CCC
02900 15 CONTINUE
03000 DO 20 I=1,2
03100 DO 20 J=1,2
03200 QR(I,J,N)=REAL(C2(I,J))
03300 QI(I,J,N)=AIMAG(C2(I,J))
03400 20 CONTINUE
03500 30 CONTINUE
03600 RETURN
03700 END
03800 C *****

```